MODEL 640

GAUSSMETER

Operating and Servicing Manual

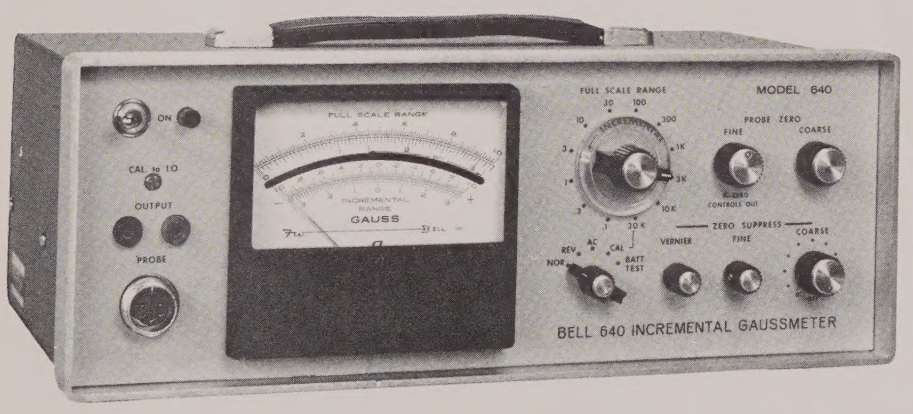
FW BELL INC.
ELECTRONIC DESIGN • DEVELOPMENT • MANUFACTURING

MP MASTER
MA30392

MODEL 640

GAUSSMETER

Operating and Servicing Manual



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SECTION I

General Description

I-1 GENERAL

The Model 640 Gaussmeter is a precision magnetic flux measuring instrument, featuring high stability, silicon, solid state construction. It has been designed especially for use with the recently developed Hall effect magnetic field probes manufactured by F. W. Bell, Inc.

A wide range of standard transverse and axial, as well as special probes, are available including the new high-linearity, low temperature coefficient Hall sensors All probes are directly interchangeable without recalibration. Factory calibration of probes is carried out with reference to a laboratory standard magnet. This standard is measured by the Nuclear Magnetic Resonance (NMR) technique and provides traceability to the National Bureau of Standards. The user can then set this calibration into the instrument by the use of the simplified high accuracy built-in CAL procedure.

Measurement range extends from 1 mG (milligauss) per scale division (100 mG full scale) to 30 kG (kilogauss) full scale. Both dc and ac fields can be measured directly using the 4½ inch panel meter. DC polarity is preserved for direction information when tracing and plotting fields. AC fields of up to 400 Hz are read out as an rms value for sine wave fields.

Also zero suppression and scale expansion controls permit incremental measurements. Any dc field reading to 30 kG may be suppressed to zero and sensitivity increased to a maximum of 1000 times, thus providing full scale output for 0.1% changes in field.

Output jacks are provided on the front panel. Output voltage is 1 volt full scale and is proportional to the field for dc and ac fields up to 400 Hz. An auxiliary output amplifier is also included providing increased output voltage and power, via rear panel jacks, to drive external equipment. To obtain high accuracy measurements, a high resolution indicating instrument such as a digital voltmeter can be connected to the output jacks. The internal calibration procedure may also be carried out using this indicating instrument to achieve the best possible accuracy.

Because of automatic change-over to battery operation, the Model 640 is ideal for fixed installations where constant, noninterrupted service is required in the event of power line failure, as well as applications requiring portability. An inexpensive, sealed, high energy dry battery is used (included with instrument on special order). Also an external dc power input is provided through binding posts on the rear of the cabinet.

Simplicity of operation, wide measurement range and high accuracy make the Model 640 a versatile instrument for measurement in the laboratory or in the field, as well as for production testing and process control.

SECTION II

Specifications

II-1 ELECTRICAL AND PERFORMANCE SPECIFICATIONS

(a) Measurement Ranges:

Measurement of static (dc) or varying (ac to 400 Hz) magnetic field strength in the range of 0.1 gauss full scale to 30,000 gauss full scale in 10 db steps as follows:

1X probes	0.1, 0.3, 1.0, 3.0, 10, 30, 100, 300, 1 kG, 3 kG, 10 kG, 30 kG
10X probes	1.0, 3, 10, 30, 100, 300, 1 kG, 3 kG, 10 kG, 30 kG

(b) Calibration:

1. Internal calibrating procedure is referenced to a standard NMR magnet traceable to the National Bureau of Standards and to probe deviation curves. Probe deviation curves in accordance with MIL-STD-793-1 (WP) Appendix A.

2. Internal calibration error does not exceed ±0.3%.

(c) Zero Suppression Range and Scale Expansion:

Any dc field up to 30 kG may be suppressed to zero. The sensitivity may then be increased in the 1,3 sequence by means of the range switch to a maximum of 1000 times (6 switch positions).

(d) Available Standard Probes:

One hundred and ten different Hall effect field probes are available to meet the challenging requirements of virtually any application. Please see the F.W. Bell Gaussmeter Probes literature sheet for probe models and prices. Consult the factory for special probe requirements.

(e) Accuracy:

1. Accuracy is the sum of accuracies of the instrument, the probe and the calibration source. The instrument accuracy is ±0.25% of full scale plus ±0.02 gauss, and the internal calibration accuracy is ±0.3% of reading. For example, the 640 gaussmeter used with the model HTB4-0608 probe calibrated with the internal calibration output; the total accuracy is ±0.25% of full scale, plus ±0.02 gauss (instrument), plus ±0.3% of reading (internal calibration), plus ±1.0% of reading (probe) for a total system accuracy of ±0.25% of full scale, plus ±1.3% of reading, plus ±0.02 gauss. The ±0.02 gauss is an instrument error. When a 10x probe is used the specification then becomes ±0.2 gauss.

2. For Incremental Measurements:
±1% to ±3% of incremental range depending on probe type and absolute field magnitude.

3. For ac Fields:
0 to 60 Hz; dc field accuracy plus ±1% of reading
60 to 400 Hz; dc field accuracy plus ±10% of reading

4. Meter Scale Tracking Error:
For dc field meter readout ±1% of full scale

For ac field meter readout 10 to 400 Hz; ±2% of full scale

5. Improved accuracy obtainable by using probe deviation curves and, at specific test points, by reference to a known calibration field.

(f) Stability:

1. Line voltage:
Error unmeasurable for ±10% line voltage changes

2. Temperature effects excluding probe influences:
Approx. ±3% of reading total over the range of −50°C to +70°C (can be removed by using internal calibration feature)

3. Temperature effects and zero temperature stability will be determined by the type probe being used. Consult the Gaussmeter Probes literature sheet for the complete specifications of the probe being used.

(g) Output Jacks Front Panel:

1. Output voltage:	1.0 volt dc FS
2. Source impedance:	1 kΩ approx.
3. Maximum ac field frequency:	400 Hz
4. Response time for full scale step input:	0.4 msec approx.
5. RMS Noise:	
0.1 gauss range	25 db below full scale approx.
0.3 gauss range	35 db below full scale approx.
1 G to 30 kG range	45 db below full scale approx.

(h) Auxiliary Output Jacks, Rear Panel:

1. Output voltage:	1.0 to 5.0 volts dc FS adjustable
2. Source Impedance:	<1Ω (current limited above 10 mA output current)
3. Maximum ac field frequency:	400 Hz
4. RMS Noise:	
0.1 gauss range	25 db below full scale approx.
0.3 gauss range	35 db below full scale approx.
1 G to 30 kG range	45 db below full scale approx.

(i) Power Requirements:

1. Line input:

Volts	105-125 V	or	210-250 V
Frequency	50-60 Hz		50-60 Hz
Current	0.038 A		0.019 A
Power	4 W		4 W

2. Battery input:
9 V to 18 V at approx. 130 mA, automatic take over in case of failure in line voltage.

3. Internal battery:
12 V dry battery, Neda 926 — Eveready 732; automatic take over in case of failure in line voltage (approx. 60 h life with continuous use)

II-2 PHYSICAL SPECIFICATIONS, CONTROLS AND CONNECTORS

(a) Front Panel:

1. Power Switch:
This toggle switch turns on the primary power to the instrument. The pilot indicates when the instrument is operating on line voltage only, no indication of battery operation is provided. See cautionary note in Section II-2b4.

2. Range Switch:
The full scale range is set by this switch in 10 db steps marked in the 1, 3, 10 sequence. It covers the range from 0.1 gauss to 30 k gauss, the full range of the instrument. Also by use of the interlocking knob and skirt the incremental scale expansion ranges are selected.

3. Function Switch:
This switch selects the mode of operation. It includes

the measurement positions for dc fields of both normal and reverse polarities, and for ac fields, the calibration position for internal CAL, and a battery test position to check the internal battery condition or external dc input voltage.

4. Probe Zero Controls:

COARSE and FINE zeroing controls are provided to balance each individual probe for zero output in the absence of a magnetic field. They will also suppress small residual fields up to approximately 30 gauss. By turning the FINE control full CCW past the switch, the zero controls can be disconnected, permitting the instrument to be used with special probes having their own zero controls.

5. Zero Suppression Controls:

Three controls COARSE, FINE, and VERNIER are used to suppress the field reading. When the COARSE control is switched from its off position, the meter circuit generates true zero center operation for simplified positive and negative deviation measurements.

6. CAL FS Control:

The screwdriver CAL control is used to calibrate the instrument using the internal calibration feature as well as calibration to a standard magnet. Calibration through external indicating instruments is also possible with connection to the output jacks.

7. Output Jacks:

The black and red front panel jacks will accept a standard double banana plug with ¾" spacing. They provide an output voltage proportional to the full-scale field, for use with external instrumentation. The polarity is positive for upscale meter readings and negative for downscale indications. The black jack is grounded to the chassis.

(b) Rear Panel:

1. Power Cord Receptacle:

The 3 pin power cord receptacle accepts the detachable power cord. The power cord is equipped with three conductors and is terminated in a three-prong plug recommended by the National Electrical Manufacturers' Association. The round pin is connected to the case and grounds the instrument case and output terminals when used with the appropriate receptacle. An adapter may be used for connection to a standard two-contact receptacle. The ground is brought out of the adapter by means of a short wire which should be connected to a suitable ground for protection of operating personnel. Only when a ground is supplied by associated equipment should this ground be unused to prevent common ground currents. The Model 640 is normally wired for 117 V nominal 50-60 Hz. To change it to 234 V, the primaries of the power transformer must be changed from parallel to series connection.

2. Fuse:

The fuse holder accepts a standard 3 AG size fuse. For 117 V operation, the fuse should be rated at 1/16 A Slo-Blo. For 234 V operation a 1/32 A Slo-Blo fuse should be used.

3. Battery Input:

An input for external battery power is provided. The yellow and black binding posts are mounted on ¾" centers to accept a standard banana plug or can be tightened directly on to prepared cable ends. The instrument will operate satisfactorily for voltages over the range of 9 to 18 volts dc at about 130 mA drain. The yellow post is positive and must be connected to the positive terminal of the dc power source. The black terminal is grounded to the instrument case.

4. Battery-Line Switch:

This switch selects the input power used by the instrument. In the line position the Model 640 will operate on the line power only (117 or 234 V). In the BATT. position the instrument will operate on internal or external battery by means of the automatic takeover feature if line power is not supplied. NOTE: WHEN THE SWITCH IS IN THE BATT. POSITION THE INSTRUMENT MUST BE TURNED OFF USING THE POWER SWITCH. Merely pulling the plug or turning off the power source to the line cord will only cause the instrument to switch over to its internal battery thus shortening the battery life. It is, therefore, suggested that the BATT-LINE switch be left in the LINE position, unless battery power or the automatic takeover feature is desired.

5. Auxiliary Output Jacks:

The black and red binding posts are at the standard ¾" spacing and provide the output from the auxiliary dc amplifier. The output voltage is adjustable from 1 Volt to 5 Volts for full scale meter reading. The output impedance is $<1\,\Omega$ and will drive up to 10 mA output current before limiting occurs for short circuit protection. The auxiliary amplifier operates only when the instrument is supplied with line power. No output will be obtained when the instrument is operated from battery or external dc power.

6. Gain:

The auxiliary amplifier gain is adjusted by the GAIN control to obtain the 1 to 5 Volt output for full scale.

(c) Overall Dimensions:

5¾" high 14" wide 10⅜" deep

(d) Weight:

Shipping 18 lbs.
Net 12 lbs.

SECTION III

Installation and Pre-Operation Adjustments

III-1 POWER

To install the battery, slide the cover back and off after removing the screws on the rear. Take off the battery mounting clamp strip near the power transformer on the top of the sub chassis. Set the battery on the chassis with its terminal socket facing the power transformer, and slide the battery can lip under the clamp strip riveted near the range switch. Replace the removed strip, clamping the battery lip to the chassis. Connect the lug attached to the blue wire to the positive battery terminal, the lug on the white wire to the negative terminal and replace the cover. The battery used is a 12 Volt dry battery NEDA number 926 — Eveready 732.

For portable operation simply disconnect the power cord (for convenience in handling) and switch to BATT position. The pilot will not operate except on line power.

(b) Power Requirements:

The Model 640 can be operated from any one of three types of power: standard ac line voltage (117 V or 234 V, 50-60 Hz, 4 watts), internal battery (dry battery 12 volt, NEDA 926 — Eveready 732) or external dc power (9 -18 volts dc at approximately 130 mA drain). The BATT-LINE switch selects either line power only in line position, or automatic battery takeover in the BATT position. In BATT position the instrument automatically selects the power source on the basis of the following criteria. Power is always drawn from the line when connected to a 117 V (or 234 V) ac source. If line power is not supplied, the instrument will operate from either the internal battery or external dc source, whichever has the higher voltage. Through this action double backup protection for power outages is possible. If the 117 V line fails, the 640 would switch over to the external dc source (it is assumed this source voltage is in excess of the 12 V supplied by the internal battery). In the event of failure of the dc source, the internal battery would take over. Because of the automatic change over, it is important that the instrument always be turned off by the power switch when the BATT-LINE switch is in BATT position. Pulling the plug or turning off the power source will only cause switching over to battery power. Also, the auxiliary amplifier will not operate from the battery or external dc input.

(c) Operation:

Before turning the instrument on, make certain the power to be used matches the voltage and frequency ratings of the gaussmeter. Battery condition or external dc source voltage can be checked prior to turning on by the BATTERY TEST position of the FUNCTION switch.

When the FUNCTION switch is set at BATT TEST position, the meter is connected to the battery and dc input circuitry. The meter monitors either the internal battery or external source whichever has the higher voltage (check above paragraph). The meter must indicate within the range marked BATT with the instrument operating. If the indication is over or under the range, then the voltage is either too high or too low for proper operation. The battery condition can also be checked prior to instrument turn on to monitor the effects of the instrument load on the source.

Connect a probe to the input socket. It is important that the plug is pushed firmly into the panel socket observing the key slot, then the clamp ring screwed on until it is snug. Check the meter mechanical zero by aligning the pointer with its image in the mirror. If necessary adjust the screw on the meter face to exactly zero reading. Set the panel controls as follows:

RANGE switch to 30 K
FUNCTION switch to NOR
COARSE ZERO SUPPRESSION switch to OFF

Turn the power switch on and allow a few minutes warm up before making any measurements or calibrating.

III-2 PROBE ZEROING ADJUSTMENTS

To reduce the residual probe output to zero, the probe zero controls are used. Set the FINE control to about mid setting. Rotate the RANGE switch counter clockwise until a reading is obtained on the meter. Adjust the ten-turn COARSE control to bring the reading near zero while reducing the RANGE setting. Below the 1 gauss range, the FINE control may also be used for better resolution of zero adjustment. Make sure the coarse zero suppression switch is OFF when zeroing the instrument.
Because the earth's magnetic field is well within the measurement capabilities of the gaussmeter, the effects of this and residual fields from other sources on zeroing must be considered, especially when measurements are made on the higher sensitivity ranges (100 gauss and below). Check Section IV-2b for details on types of zeroing.

III-3 CALIBRATION PROCEDURE

(a) General:

The calibration procedure should be carried out before using the instrument for magnetic field measurements. It should also be checked whenever probes are changed. Allow sufficient time for warm up stability before calibrating.

Calibration consists of adjusting the element control current in accordance with the sensitivity of the probe and gain of the amplifier by using one of the following two methods.

(b) Calibration Against Internal CAL Signal:

Calibration against the internal CAL signal is a simple, easy to apply technique which will, in most cases, produce the best possible accuracy. Calibration against a reference magnet can produce better accuracy only if the reference magnet is better than ±0.3% and the probe linearity curve is known.

The accuracy and stability of the internal calibration signal is dependent only on precision, high stability resistors and the initial factory calibration. Therefore, long-term, high stability is a characteristic of this technique. It is recommended that, unless a very high accuracy reference magnet and probe linearity data is available, the internal calibration be used.

After the instrument and probe have warmed up (5 min.) set the RANGE switch to the 30 K range and the FUNCTION switch to CAL. Adjust the CAL screwdriver adjustment to obtain a full-scale (1.0) meter reading (or a full-scale reading (1-V range) on any external readout instrument connected to the output jacks). The gaussmeter and probe are then calibrated to the accuracy shown in the specifications under Section II-1d. Because the adjustment is the same for all probes, it is not necessary to recalibrate each time the probe is changed.

(c) Calibration Against a Reference Magnet:

Greater accuracy can be achieved by the use of a reference magnet, provided the magnet has an accuracy

of better than ±0.3% and the probe linearity data is known. Deviation error curves are available with most F. W. Bell Gaussmeter Probes at a nominal charge. Curves may be obtained at time of purchase or after purchase by returning the probe to F. W. Bell.

The linearity curve is machine drawn using a precision electromagnet and is an error plot for the particular probe measured. It is a plot of the probe's deviation from the true value over the measurement range.

The curve is plotted with actual flux density along the horizontal X-axis and the deviation from true value, in gauss, vertically along the Y-axis. Thus, the locus of deviation errors of, for example, 1% of the actual field, will appear on the sheet as a diagonal line passing through the 1% of value points.

To calibrate using the linearity curve and reference magnet, the following procedure should be used. It is important that the absolute zero procedure be carried out prior to calibration by this method (see Section IV-2b). The Hall probe must be carefully positioned in the field to the correct location and orientation to respond to the correct field magnitude (maximum reading) without alignment errors. Set the RANGE switch to the proper range giving the maximum on-scale meter reading (or external readout instrument reading). Add the reference magnet specified field, in gauss, to the deviation, in gauss, read from the linearity curve at that flux density. Adjust the CAL FS control to obtain a reading equal to the sum. For example if the reference magnet were specified at 9.8 kG and the deviation from the curve was −10 G, then the adjustment would be made to obtain 9.79 kG on the meter (or external readout instrument if used).

(d) AC Field Calibration:

No separate ac field calibration is necessary. If the gaussmeter has been properly calibrated for dc fields, then the ac signal at the output jacks for ac fields will be calibrated to the same accuracy, with the exception of the effects of the frequency response.

The frequency response is essentially flat to 60 Hz. From 60 to 400 Hz the response may vary up to about 10% or 1 db. Above 400 Hz the response rolls off rapidly and the gaussmeter is unusable for measurements. The reduced accuracy for ac measurements indicated in the specifications is for the panel meter readout and is a result of the additional errors introduced by the ac detector circuit. These errors are not present in the signal at the output jacks and, therefore, higher accuracy ac measurements can be obtained through the use of high accuracy ac external readout instruments.

SECTION IV

Operating Instructions

IV-1 THE PROBE

(a) General:

There are ten standard probes available for the Model 640 Gaussmeter using an improved design having a 7/16 inch diameter aluminum handle. This handle tube can be used for clamping the probe to a support or holding fixture. The standard probes are divided into two basic categories depending on field direction response: transverse and axial. Figure 1 shows the shape and size of the various probes. In addition special probes of all types can be designed to meet unusual requirements. Contact F.W. Bell, Inc. for information and quotations.

Although the probes are quite rugged, reasonable care should be exercised in handling. Avoid excessive shock, pressure, bending or otherwise straining the element mount. The element can be fractured if over-stressed since it is brittle as well as rigid.

Note that probes are supplied for various flux measurement ranges. For accurate readings, make sure that the probe is being used within its specified range of measurement. When 10 kG probes are used for measurement higher than 10 kG, increased deviation from linear response will result. Such measurements would not be accurate as absolute readings, but might be useful where only a relative indication is sufficient.

AXIAL PROBE:

TRANSVERSE PROBE:

AXIAL FLEXIBLE:

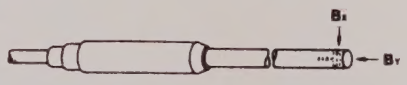
TRANSVERSE FLEXIBLE:

REQUIREMENTS	SUGGESTED PROBE CONFIGURATION
general use; durable	HEAVY DUTY
measure flux density in awkward places	FLEXIBLE
measure flux density in a small gap	STANDARD DUTY (laboratory use)
measure homogeneous fields* from 0.001 G to 2 G; high sensitivity	MAGNAPROBE
high linearity; low temperature coefficient; accurate field measurement to 150 kG	10X PROBE**
multi-axis field measurement	2-AXIS PROBE; 3-AXIS PROBE
withstand low temperature (down to −270°C)	CRYOGENIC
monitor instantaneous difference between two field points; field mapping, homogeneity testing	DIFFERENTIAL

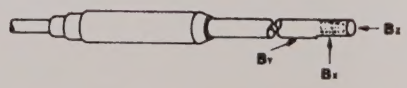
*averages flux density along 9'' probe length

** 10X probe sensitivity is one-tenth standard probe (1X) sensitivity

TWO-AXIS PROBE:



THREE-AXIS PROBE:



(b) Probes:

A transverse probe, as its name implies, has its flux response direction transverse to the probe handle axis. Notice that the flux vector is shown entering the Hall element at right angles (90°) to the plane of the element surface. This is the direction of maximum Hall output. Magnetic fields entering the element at some other angle to the surface cause an output response proportional to the 90° component. That is, the response will be $B \cos \Theta$ where B is the field magnitude and Θ is the angle of the field to the perpendicular.

The axial probe has its flux response direction parallel and on axis with the probe handle. The element is encapsulated on the end of the probe extension tube. If the field is directed at some angle with respect to the probe axis, only the axial component will produce an output, just as the transverse component will only produce output for transverse probes.

IV-2 MEASUREMENT PROCEDURE

(a) General:

After completing the pre-operational adjustments described in Section III, the unit is ready for operation. Allow sufficient time after turn-on for the unit and probe to reach operating temperature (5 min.). It is advisable to position the probe securely in the measurement position using a probe holding fixture or clamp whenever possible. It may be difficult or impossible to obtain a good field measurement unless the probe is stable. The probe may be hand-held only when operating on the higher ranges (100 gauss and above) and the field is fairly uniform over a reasonable area.

When using F. W. Bell probes, errors in readings due to temperature changes at the probe can be corrected by using the mean probe temperature coefficient and computing corrections (see Section IV-2e).

(b) Zeroing:

Since Hall probes do not have exactly zero output in a zero field, it is necessary to electrically zero the gaussmeter before taking any readings. Two types of zeroing are possible: Absolute and Relative.

Absolute zero is required whenever it is desired to know the actual field at the probe. As its name implies, the controls are adjusted to give zero reading at zero field. Because of the ever-present earth's field and possible stray fields from other sources, it is necessary to shield the probe to achieve zero field during the time of adjustment. The zero gauss chamber accessory available for the probe is a dual mu-metal can assembly which effectively shunts all external stray fields around the internal volume. By slipping this chamber over the probe end, a zero field condition is created at the element and the gaussmeter is adjusted for zero output by the procedure of Section III-2.

Use care not to place the zero gauss chamber in a strong magnetic field. Also, it must not be allowed to come into direct contact with a magnet, since it may become slightly magnetized and will not provide a true zero. It can be de-magnetized by slowly passing the chamber through the ac field of a demagnetizer coil carrying ac line current.

The zero chamber provides a "true" zero reference. Readings taken in the presence of the earth's ambient field will generally have the earth's field, or some component of it, included in the reading of the unknown field. Thus, it may be necessary to subtract the ambient field reading from the total to obtain the value of the unknown field. To avoid this subtraction, the method of relative zeroing may be used. In this case the gaussmeter is zeroed after the probe has been placed in the measurement position, without using the zero gauss chamber. The ambient field is therefore zeroed out electrically in the gaussmeter. This method is successful only if the field to be measured can then be presented to the probe without changing the probe position with relation to the ambient field, and without altering the ambient field at the probe during measurement. The change in field at the probe will be measured by the gaussmeter as an absolute value, and the ambient field excluded from the reading. Since the probe position must remain fixed during the measurement, this method is not always practical. The zero controls are capable of suppressing residual fields of up to about 30 gauss.

Since the ambient field seldom exceeds 1 gauss, the precautions mentioned above apply only when measuring fields less than 500 or 1000 gauss. If a magnet has an iron structure which will modify the ambient field by its presence, it may be necessary to take several measurements in different orientations with respect to the earth's field. Obtain the two extreme values and use the mean value between these as the correct value.

The zero adjustments should be checked frequently, if possible, during the course of a measurement particularly when using low ranges. If a change in temperature occurs at the probe, rezeroing may be necessary. Zero drift versus temperature is small and is not compensated. The zero controls are never used to shift the calibration of the gaussmeter.

(c) Calibration Check:

A calibration check can easily be made at any time during a series of measurements without disturbing the probe by using the internal calibration procedure. Refer to Section III-3b on calibration.

(d) Operation:

1. DC Fields:

For absolute dc field measurements set the FUNCTION switch to NOR, the Coarse zero suppression switch to off, and the RANGE switch to a range higher than the expected field. The RANGE switch is a dual concentric switch with an interlock between the knob and skirt. For all absolute measurements the knob pointer should align with the black index mark on the skirt. If it does not, rotate the knob until they do align. The knob will snap back and interlock with the skirt. The knob and skirt will then rotate together.

With a field at the probe the meter will read up scale if the direction aligns with the probe sensitivity direction as described in Section IV-1b. If the meter tends to read below zero, the field direction is reversed and the probe can be turned over or the FUNCTION switch can be set to REV. Switch the RANGE switch to a range which gives the greatest onscale meter reading. To read the magnitude of the field, adjust the probe for maximum reading. If a desired direction component of a field is to be measured, set the probe so that the sensitive direction aligns with the desired component. The meter will then read the component.

The measured value is determined from the meter reading and RANGE switch setting. All absolute measurements are from the black meter scales. The RANGE switch indicates the full scale value in gauss of the meter reading. For example, if the RANGE switch is set on 10 kG and the meter reads 1.0 on the black 0 to 1 scale, then the field is 10 kG. If the meter reading were 0.5, then the field would be 5,000 gauss. When the RANGE switch is on a 3 range (3 kG, 300 kG, etc.,) the meter reading is taken from the lower black scale. A RANGE switch setting of 300 G and a meter reading of 2 on the lower scale would indicate 200 G.

2. Incremental DC Fields:

To measure a small deviation in a larger field, zero suppression and scale expansion is used. Obtain an absolute field measurement as described above; and

with the meter reading the field, advance the COARSE suppression switch. Adjust the switch until a near zero meter reading is obtained on the RED zero center incremental scale. Adjusting the FINE suppression ten turn control and VERNIER control permit a precise zero reading. At this point the initial absolute field reading has been suppressed to zero and the meter circuit has switched for zero center operation. To expand the scale for incremental measurements the RANGE switch is used. Pull the knob out to unlock it from the skirt and rotate it counter clockwise for more sensitivity. The skirt must not rotate when scale expansion is applied. Trimming the FINE and VERNIER Suppression controls will achieve more accurate zero suppression.

To determine the incremental deviation in the absolute field, the RANGE switch knob and red meter scales are interpreted. The RANGE switch indicates the incremental range and the meter indicates the value. If the RANGE switch knob were on the 10 G range and the meter read +0.6 on the upper red scale the absolute field would have increased 6 gauss. If the switch were on the 30 gauss range and a reading of −2 was observed on the lower red scale, then the field had decreased by 20 gauss.

The maximum scale expansion permissible is limited to 6 switch positions or 1000 times the absolute range. This maximum is shown by the red index marked on the skirt. Because of the interlock, attempting to increase the expansion beyond 6 switch positions will cause the skirt to rotate and produce an off scale meter deflection.

3. Operating Example:

(a) Deviations in an 8.5 kG field are to be measured by the following steps.
(b) The absolute field is measured and produces a meter reading of 0.85 on the upper black scale when the RANGE is set on the 10 k range.
(c) The COARSE ZERO SUPPRESS switch is advanced from OFF until the meter reads just above zero on the red incremental scale.
(d) The FINE SUPPRESSION control is advanced until a better zero is obtained.
(e) To apply scale expansion the RANGE switch is now pulled out unlocking it from the skirt and rotated counterclockwise to the 10 gauss range for maximum expansion.
(f) The FINE and VERNIER SUPPRESSION controls are trimmed to obtain a precise zero.
(g) The instrument is now operating in its incremental mode and an increase or decrease of 10 gauss in the 8.5 kG field will indicate as a plus or minus full scale meter reading on the red zero center incremental scale.

4. AC Fields:

For ac field measurements the FUNCTION switch is set to the AC position. The direction response of the Hall element is the same as for dc fields and alignment of the probe and field direction must be considered as described above. The meter is calibrated to read the rms value for a sine wave flux. If the field has much harmonic content, the reading will not be the true rms value. The ac detector is an averaging type circuit. For non-sinusoidal wave forms within the 400 Hz band width, the average value can be obtained by multiplying the meter reading by 0.9. The meter reading and RANGE switch setting determine the field measurement just as in the case of dc field measurements. The frequency response for ac field measurements using the meter is from 10 Hz to 400 Hz. Frequencies outside this range will not indicate correctly on the meter. The amplifier system, however, is capable of operating from dc to 400 Hz, and measurements below 10 Hz can be made by using a dc coupled oscilloscope or low frequency voltmeter connected to the output jacks. AC fields cannot be read incrementally because the suppression controls can suppress only steady dc fields or the dc component of an alternating field.

5. Overloading:

Because the amplifying system used in this gaussmeter will amplify ac and dc fields simultaneously, attempting to measure a small ac field in the presence of a large dc field or a small dc field in the presence of a large ac field can overload the amplifier and cause erroneous measurements. Also when making incremental dc field measurements any ac ripple on the field can cause difficulty especially at high scale expansion. When measuring dc fields absolute or incremental, especially on the sensitive ranges, check for ac fields by switching to AC. If an ac reading exceeding ⅓ of full scale on the range in use is encountered, the dc measurement may be in error. Switch to the next less sensitive range. If the dc reading has not changed, the ac component of field has not overloaded the amplifier. If at any time there is a discrepancy between ranges of more than 2% of full scale on the meter, there is the possibility that the amplifier is being overloaded, and the above tests should be made. Also when making ac measurements the component of any dc fields must be kept below 10% of the range in use. This can usually be accomplished by using the zero controls to suppress the small earth's field or residual fields. Relative zeroing as described in Section IV-2b will remove any dc fields up to 30 gauss. If this is not sufficient to prevent overloading the suppression controls may be used or a less sensitive than normal range can be used to prevent the undesired dc field from overloading the amplifier.

(e) Temperature Effects:

All Hall probes exhibit a certain temperature coefficient. This value can be found in the specifications for the type of probe being used. If the probe temperature is known relative to the temperature at which it is calibrated, correction for the temperature effects can be applied to the measurement readings. It is recommended, however, that if the probe temperature is not known or rapid variations are expected, they will give greater accuracy in the presence of temperature variations than conventional probes with calculated corrections, and the stability will allow high resolution measurements with external indicating devices.

To correct for temperature influence, the following formula can be used:

$$B_a = B_g \left(1 - (t_{pm} - t_{pc}) \frac{t_c}{100}\right)$$

where: B_a = the actual value of the field being measured (unknown).
B_g = the value of the field as indicated by the gaussmeter.
t_{pm} = temperature of probe at time of field measurement in degrees C.
t_{pc} = temperature of probe at time of calibration in degrees C. (25°C when using internal CAL method).
t_c = temperature coefficient of probe in % per degree C.

This formula will correct the gaussmeter reading (B_g) to give the actual field (B_a) for measurements under different temperature conditions. Notice that if the t_c is a negative number, the minus sign must be carried into the formula. Note also that the temperature effects are on the absolute field readout of the instrument. Therefore when operating incrementally the temperature effects are correspondingly increased. The temperature change which produces a 0.1% change in the absolute field measurement will produce a full scale deviation when 1000 times scale expansion is used. Care should be taken, when making high sensitivity incremental measurements, to minimize temperature changes at the probe. Also temperature compensated probes will considerably improve the stability of the incremental measurements.

IV-3 OUTPUT JACKS

(a) Front Panel Jacks:

The output jacks provide the electrical output of the instrument. They provide the highest resolution, stability and accuracy capability of the Model 640 The red and black jacks are at the standard ¾" spacing to accept a dual banana plug. The red jack is the output which

is positive for upscale meter readings and negative for below zero meter indications. The black jack is the common and is grounded to the gaussmeter case. The voltage available is 1 volt for full scale meter reading and directly proportional to the instantaneous field at the probe, for frequencies from dc to 400 Hz. The source resistance is 1 kΩ and the jacks can be loaded with any desired impedance, thus permitting connection to any type of external indicating instrument.

High resolution and accuracy can be achieved by connecting a digital voltmeter to the output jacks. Set the digital voltmeter to its 1 volt full scale range and calibrate the gaussmeter as described in Section III-3 using the DVM as the readout instrument. By using only the decade ranges (0.1, 1, 10, etc.) on the gaussmeter, the DVM will read directly in gauss with full scale being equivalent to the full scale range setting of the gaussmeter.

Because of the stability (especially with the temperature compensated probes) and back up battery power, the 640 is ideally suited to long term field monitoring applications. By connecting a chart recorder or other recording device to the output jacks, the model 640 becomes a gauss-to-volts transducer for these applications.

(b) Rear Panel Auxiliary Output Jacks:

The rear panel output jacks are connected to the output of the auxiliary dc-to-400 Hz amplifier. The red jack is the output and black is ground. The output voltage is adjustable by means of the GAIN control from 1 volt to 5 volts at full scale. The output impedance of these jacks is less than 1Ω and short-circuit-protected by current limiting above 10 mA. These jacks may be used to drive lower sensitivity indicating equipment such as galvanometer pen recorders, electronic gates, etc. Because the auxiliary amplifier operates only when line power is supplied, the front panel jacks must be used for electrical output when the instrument is operated from battery power.

IV-4 PROBE DEVIATION CURVES

Machine drawn error curves can be obtained at a nominal charge for any probe, at the time of purchase or by returning the probe to F.W. Bell, Inc. The use of the probe deviation curves for calibrating is described in Section III-3c.

To use the curves to correct the field measurements, locate the field value, as indicated on the gaussmeter, on the horizontal axis of the curve. Use the right half of the curve for positive or normal direction fields and the left half for negative or reverse direction fields. Read the deviation in gauss from the curve. If the deviation is positive (above the axis), the probe output is high and the error is to be subtracted from the indicated value. Negative probe deviations are added to the readings to obtain the corrected value.

IV-5 TOTAL FLUX DETERMINATION

The Hall-effect sensing device is inherently responsive to magnetic flux density and not to total flux lines. It is not dependent on rate-of-change of flux as is a search coil.

From the basic definitions:

$$\text{GAUSS} = \text{WEBERS PER SQUARE METER} \times 10^4$$

$$\text{GAUSS} = \text{LINES PER SQUARE INCH} \times .1550$$

The Hall probe is equally useful in homogenous, uniform low-gradient fields and in high-gradient fields, although the low-gradient fields are capable of more accurate measurement. If the flux density is uniform and unidirectional over a given area, the total flux through the area is found by multiplying by the area in question.

$$\text{(Flux) WEBERS} = \text{GAUSS} \times \text{SQ. METERS} \times 10^{-4}$$

$$= \text{GAUSS} \times \text{SQ. CM} \times 10^{-8}$$

$$\text{(Flux) LINES} = \frac{\text{GAUSS} \times \text{SQ. INCHES}}{.1550}$$

When the Hall generator is moved in a plane, the component of flux normal to this plane will be indicated.

If the field varies in magnitude over the area, it is necessary to integrate the values of flux density over the area in question. The flux density value indicated by the Hall probe output is the effective value over the active area of the Hall generator. Fortunately, the Hall-Pak devices have extremely small active area. The standard transverse probe active area is only about 0.003 square inch, and sensitivity is essentially uniform over this area.

Absolute measurements of total flux of magnets having odd shape or high length-to-diameter ratio are best made using the search coil and standard fluxmeter methods. In many cases, however, valuable data are obtained by air-gap flux density measurements when the magnet is mounted in its working structure. Also, accurate comparison data can be obtained on almost all magnets of various sizes and shapes using a Hall probe to measure pole face density in comparison to a magnet selected as a standard of reference.

SECTION V

Theory of Operation

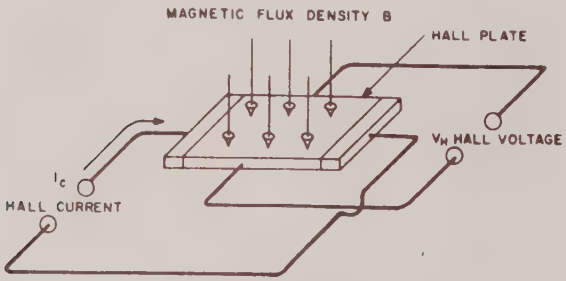
V-1 GENERAL

The basic principle of magnetic flux measurement used in the Model 640 Gaussmeter may be described as a flux-modulated carrier-amplifier system. A locally generated ac carrier signal is fed as an exciting current to the Hall element. The flux to be measured at the probe modulates the ac carrier within the Hall sensor, producing an ac output voltage which is accurately proportional to flux density. The flux modulated ac carrier output is amplified by the carrier amplifier and then restored to a flux proportional voltage by the synchronous demodulator without loss of polarity (field direction) information. The demodulator drives the panel meter and output jacks.

If the field is time varying the demodulated output will be instantaneously proportional to the flux waveform up to 400 Hz. A separate ac detector is connected to the synchronous demodulator output to produce a dc meter reading for ac fields.

V-2 THE SENSING ELEMENT

The Hall-Pak generator used for magnetic flux sensing is a semiconductor device operating on the Hall-Effect principle. It consists of a thin rectangular wafer of high-purity indium arsenide with 4 leads attached, Fig. 2.



THE HALL GENERATOR

FIGURE 2

The application of control current I_c to the Hall generator results in a flow of charge carriers through the semiconductor material in the direction of its long dimension. When the Hall generator is placed in a magnetic field, the Lorenz force, acting on the moving charges, deflects them at right angles to the direction of their motion through the Hall plate. This is the same force that deflects the electron beam in a cathode ray tube.

The resulting build-up of charge carriers along the sides of the wafer produces the Hall voltage, and this voltage appears as an output at connections made on each side of the element. Hall voltage V_H is directly proportional to the flux density B and to the magnitude of control current I_c.

$$V_H = K_H (B \times I_c)$$

The three factors V_H, I_c and B are mutually perpendicular. If the magnetic flux vector B is not perpendicular to the face of the Hall generator, the Hall output will be proportional to the component of B that is perpendicular to the element. The constant of proportionality K_H is called the Hall sensitivity constant, and is approximately 0.075 volt per kG-ampere for 1X probes, and 0.0075 volt per kG-ampere for 10X probes.

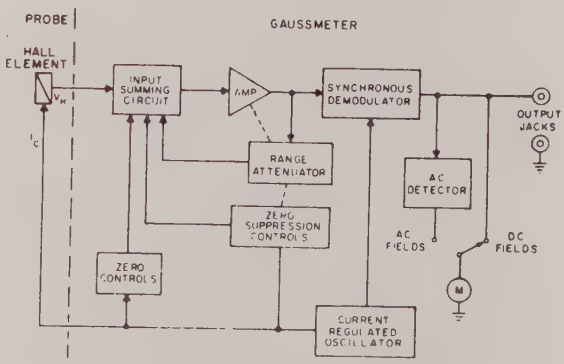
V-3 CIRCUIT OPERATION

(a) Introduction:

The section describes the overall gaussmeter circuit operation. Check the simplified block diagram, Fig. 3; the detailed block diagram, Fig. 4; and the complete schematic, Fig. 5. Because of the unique circuit design it is recommended that a thorough understanding of the circuit operation be acquired before attempting to service or adjust the instrument.

(b) General Operation Description:

The Hall element is supplied with exciting current, I_c, from the 5 kHz current regulated oscillator, Fig. 3. The Hall output voltage, V_H, (the flux modulated carrier signal) is connected to the gaussmeter input summing circuit. At the summing circuit the zeroing voltage from the zero controls, the zero suppression voltage and the feedback voltage from the range attenuator are subtracted from the Hall output. The difference is amplified to produce the proper output depending on the range in use. The amplifier output feeds the range attenuator which sets the proper gain, controls, and stabilizes the amplifier by means of overall loop feedback. The zeroing voltages are obtained from the zero control circuits which adjust and attenuate a sample of the control circuit. The amplifier output feeds the synchronous demodulator which converts the amplified flux modulated carrier into a flux proportional output voltage. This output voltage then feeds the output jacks and meter for dc fields. For meter reading of ac fields, the ac detector is connected between the output and meter to convert the flux proportional ac output to dc.



MODEL 640
SIMPLIFIED BLOCK DIAGRAM

FIGURE 3

(c) Detailed Operating Description:

1. General:

The circuit is divided into 6 basic functional groups: the probe, the input section, the amplifier section, the detector and meter, the current regulated oscillator, and the power supply. Fig. 4. the complete block diagram, shows all essential operating functions; however, various circuit details have been eliminated or altered for clarity. Check the complete schematic Fig. 5 for all circuit details.

2. The Probe:

The excitation current is supplied to the element through pins A and D of the probe plug and the Hall output voltage is connected to the gaussmeter through pins C and E. A calibrating resistor, mounted in the probe plug across pins B and F, accurately programs the amplifier gain to the proper value according to the sensitivity of the particular Hall element.

3. Input Section:

The FUNCTION switch, in the operating positions, connects the Hall voltage output directly to the input transformer. The primary winding of the transformer is split with the two resistors R3 and R4 inserted between the halves and grounded, to form a balanced to ground summing point. The input transformer, in effect, sees the difference between the Hall input voltage and the summing point voltage. The summing point voltage consists of the feedback voltage from the range attenuator, the zero injection voltage from the zero controls, and the zero injection voltage from the zero suppression controls. Also shunting across R3 and R4 is the calibrating resistor or network in the probe. By varying this CAL resistor, the feedback factor of the range attenuator is adjusted, and the overall amplifier gain is programmed very accurately for proper calibration.

The zero injection voltage from the zero controls is obtained by sampling the element control current. The controls adjust and attenuate the voltage obtained from the current sensing transformer T201. Because the Hall sensor zero and residual field output is proportional to control current, and the zero injected voltage is derived from the current, cancellation of the zero field voltage is maintained even under the condition of current changes. Highest possible zero stability is, therefore, achieved.

The zero suppression controls also adjust and attenuate a sample of the element excitation current and feed it to the summing point across R3 and R4. This voltage then subtracts from the element output produced by the field, thus suppressing the zero. The attenuator consists of two sections. The COARSE SUPPRESSION switch consists of two bridged tee attenuator sections switched in 10% steps from zero to full output. The FINE control and attenuator operates in parallel with this dual bridged tee to provide smooth control between the steps of the COARSE control. The VERNIER provides very fine control for precise zeroing. The output of this attenuator and the FINE and VERNIER controls feed into the 4 section attenuator connected to the range switch ring knob. This attenuator sets the range of the suppression voltage to correspond to the absolute range in use. During scale expansion this attenuation does not change thus maintaining the zero suppression.

During calibration the FUNCTION switch connects the internal gain programming resistor R2 in place of the resistor in the probe and disconnects the zero controls. The internal gain calibrating resistor programs the gain to the proper value for the nominal Hall element sensitivity

The input is switched from the probe output to a resistor R1 in series with the control current to the element which is calculated to produce a voltage equal to the nominal element output in a 31.622777 kG or $10\sqrt{10}$ kG field (full scale (1.0) on the 30 kG range). The CAL control which varies the current is then adjusted to give full scale output (1.0). By this method of adjustment, any amplifier gain errors are compensated for by the control current adjustment, providing an internal calibration which is dependent only on the accuracy of the resistors R1, R2, R3, and R4 and the initial factory calibration which determines the probe calibrating resistor.

4. Amplifier Section:

The input transformer secondary connects to the first attenuator which operates on the middle three ranges (30, 100, 300G) to keep the input below the saturation level of the first amplifier. The first amplifier is a two-stage plus emitter follower direct coupled circuit with ac and dc feedback for stabilization. The voltage gain is 46 db at the carrier frequency and is attenuated at all other frequencies by the synchronous filter. The filter consists of two transistor switches operating in synchronism with the carrier frequency and two RC filter networks inserted in the feedback network of the amplifier section. The action is that of a very narrow band pass filter synchronized to the carrier frequency. Very high rejection of noise and spurious signals is achieved, and easy alteration of band width is possible by changing the RC network.

The first amplifier feeds the second attenuator which operates on the first 4 ranges (0.3, 1, 3, and 10 G) to prevent overloading the second amplifier. The second amplifier is identical to the first and gives another 46 db gain at the carrier frequency with synchronous filtering. Again the output feeds an attenuator which operates on the highest 4 ranges (1 K, 3K, 10K, and 30 kG).

A third 46 db filter amplifier feeds a dual complementary unity gain output amplifier which drives the output transformer. The output transformer has two center tapped outputs, one for the range attenuator and the other for the synchronous demodulator. The 500 Ω range attenuator connects the output back to the summing point across R3 and R4, giving about a 40 db feedback factor for high gain stability, high input impedance and high linearity.

The range attenuator consists of (4) 500 Ω "O" pads switched in a binary sequence (1, 2, 4, 4) to give 10 db steps to 110 db total attenuation. Also connected to the range attenuator switch are the three dividers between the amplifiers which also operate in 10 db steps to maintain the 40 db feedback factor.

5. Detector and Meter:

The synchronous demodulator consists of two switching transistors driven by the sync. signal from the carrier oscillator. When the FUNCTION switch is in the REV position, the output transformer signal polarity is reversed which changes the polarity of the demodulated output. The output jacks are connected to the demodulator through the ripple filter R53, R54 and C1.

The meter is connected through the R49 calibrating resistor for dc fields and CAL. For ac fields the meter is switched to the ac detector. The detector consists of a diode bridge connected in the current feedback loop of a two-stage amplifier. High linearity ac to dc conversion of the flux proportional demodulated output is, therefore, achieved. For battery test, the meter is switched to the attenuator formed by R50 and R51 fed from the diodes which select the higher voltage; battery or external dc input.

6. Current regulated Oscillator:

The element control current is supplied from the 5 kHz square wave current oscillator. The oscillator consists of an astable multivibrator driving a pair of push-pull output transistors into the output transformer. The output current amplitude is controllable by a dc control signal. This control signal is obtained from a dc amplifier with positive feedback for infinite gain. The input to the amplifier is then obtained from the current sensing transformer T201 through a zener reference diode. By means of this feedback loop, the control current is regulated to obtain a high stability element excitation current.

The current is adjustable by the CAL control which varies the secondary load on the current sensing transformer. The current sensing transformer also feeds the zeroing controls.

7. Power Supply:

The power to all circuits is supplied from the power regulator which provides +7 V dc. The input to the regulator is obtained from the power transformer, rectifier diodes, and filter capacitor. At 117 V ac line voltage (or 234 V) the unregulated dc voltage is about 20 V. Because the internal battery and external input are less than 20 volts, the diodes in series with these power sources are reverse biased and no power is drawn from them. Therefore, if the ac line power fails, the external or internal battery will automatically take over, when the LINE-BATTERY switch is in the Batt. position.

SECTION VI

Maintenance

VI-1 INTRODUCTION

This section contains the necessary instructions and diagrams for maintaining the Model 640 Gaussmeter. In addition to the schematic diagram, a block diagram is provided to aid in troubleshooting. Also included are circuit board layouts.

Repair and adjustment of the instrument should be attempted in the field only where adequate test equipment and qualified personnel are available. Refer to the warranty page for the procedure to be followed should factory repair service be required.

VI-2 TEST EQUIPMENT

The following test equipment is required to test and adjust the Model 620 Gaussmeter:

(a) A high impedence dc voltmeter having 2% or better accuracy.

(b) A high impedence ac voltmeter having 2% or better accuracy. The meter should be an average reading circuit calibrated in rms for sine wave voltages. A true rms reading meter will not read the same voltages because of the square wave forms.

(c) A high quality oscilloscope having response to dc.

VI-3 COVER REMOVAL

CAUTION — Always disconnect power cord from power line when removing or replacing cover.

To remove cover, it is only necessary to remove the four screws on the rear of the unit and slide the cover off.

VI-4 PERFORMANCE TESTS

This paragraph describes a rapid overall series of tests for proper operation. If difficulty is encountered, proceed to paragraph VI-5.

(a) Preliminary:

1. Make sure power cord is properly connected and fully seated in the receptacle on instrument.
2. Check fuse for burnout. Correct fuse is:
1/16 A 3AG slow-blow for 117 V operation.
1/32 A 3AG slow-blow for 234 V operation.
3. Inspect the probe for damaged element or cable, or poor plug contacts.
4. Check meter mechanical zeroing.
5. Set controls as follows:
RANGE switch to 30 K
FUNCTION switch to CAL
COARSE SUPPRESSION to OFF
6. Plug probe securely into front panel socket.

(b) Line Current Check:

1. Connect to rated power source. Line current after warmup should be:
at 117 V line, 0.038 A nominal
at 234 V line, 0.019 A nominal

(c) Calibration Check:

1. With the instrument operating the meter should read near full scale.
2. Adjust the CAL control. A full-scale meter reading should be obtainable with a small adjustment leeway on each side.
3. Voltage at output jacks should be 1 volt dc. for full-scale meter reading.

(d) Zero Test:

1. Switch the FUNCTION switch to NOR. Adjust the zero controls to obtain zero meter reading with the zero gauss chamber over the probe on the 0.1 G range.
2. Set the RANGE switch to the 100 gauss range.
3. Rotate the COARSE ZERO control over its range, switching the FUNCTION switch to REV when below zero meter indications are obtained.
4. A total of about 60 gauss adjustment range should be obtained.

(e) Detector Zero Test:

1. Again zero the instrument on the 0.1 G range with the FUNCTION switch at NOR.
2. Switch to the 30 K range and measure the dc voltage at the output jacks. It should be less than 1 mV dc.

(f) Noise Check:

1. With the zero gauss chamber in place and a good zero on the 0.1 G range, switch the FUNCTION switch to AC.
2. The residual meter reading should be less than 10% of full scale on the 0.1 G range.

(g) Suppression Circuit Check:

1. With the zero gauss chamber in place and a good zero on the 0.1 G range, set the FUNCTION to NOR, the range to 10 kG, and the FINE SUPPRESS and VERNIER to full counterclockwise.
2. Advance the COARSE SUPPRESS from its OFF position to its 0 position.
3. The meter should move from zero on the black scales to zero on the zero center red scales.
4. Continue to advance the COARSE SUPPRESS control. The steps should be approximately 10% of full scale in the negative direction on the upper red scale.
5. The FINE SUPPRESS control should produce smooth continuous variation between the 10% steps of the COARSE control.

VI-5 ADJUSTMENT AND ALIGNMENT PROCEDURE

IMPORTANT — None of the adjustments described in this section should be disturbed unless the instrument is malfunctioning and the tests indicate adjustment is necessary. These tests and adjustments are designed to assure correct overall performance. Before making any adjustments, the Model 640 should be turned on for at least one hour with the cover in place. Line voltage should be at rated value (117 V or 234 V).

(a) Meter Mechanical Zero:

For this test the meter terminals may be short-circuited with a jumper wire, or the instrument turned off, allowing a few seconds for complete discharge of all capacitors. The instrument must be the normal horizontal operating position. Read the meter accurately by aligning the pointer with its image in the mirror. If necessary, adjust the screw on the meter face to bring the pointer to exactly zero.

(b) Power Supply Voltage Adjustment:

Adjust R224 on the oscillator power supply circuit board to obtain +7 V dc between terminal 206 and ground.

(c) Oscillator Symmetry Adjustment:

Connect the oscilloscope to terminal 212 and ground on the oscillator power supply board and adjust R211

on that board for a symmetrical square wave. Use the largest scope trace possible.

(d) Detector Zero Adjustment:

Place the probe in the zero gauss chamber and zero the gaussmeter down to the 0.1 gauss range. Switch to the 30K range and adjust R151 on the amplifier board to obtain zero dc voltage at the output jacks. Use the highest sensitivity voltmeter range.

(e) Meter Calibration dc:

Switch the RANGE switch to 30 K, the COARSE SUPPRESS control to OFF, and the FUNCTION switch to CAL. Adjust the CAL F.S. control to obtain exactly 1 volt at the output jacks using the dc VTVM. Adjust the potentiometer R52 on the circuit board on the back of the meter to obtain exactly full-scale meter reading. The more accurate the dc VTVM used, the more accurate the setting of 1 volt output for full scale.

Switch the FUNCTION switch to NOR and zero the instrument to the 0.1 G range. Set the RANGE switch to 30 K, the FINE SUPPRESS and VERNIER to maximum counterclockwise, and COARSE SUPPRESS to 0. Adjust R58 on the meter circuit board to obtain exactly zero on the zero center red scale.

(f) Meter Calibration ac:

With the COARSE SUPPRESS switch OFF, connect the ac VTVM to the front panel output jacks and place the probe in a source of ac field about 60 Hz. This field can be generated by any method: The external field of a power transformer, a coil operating from line power, etc. No particular accuracy is necessary, but a field of about 10 gauss would be easy to use. Mount the probe in the field using a holding fixture. Zero the dc field, if any, using the zeroing controls and on the range which gives about a 1 volt reading on the ac VTVM. Adjust the probe to provide exactly 1 volt on the ac VTVM and adjust R165 on the amplifier board to obtain exactly full scale on the Gaussmeter meter with the FUNCTION switch on AC. It is recommended that the ac field used be above the 1 gauss range because stray dc fields will not be as apt to cause difficulty. Be sure there is no dc field reading in the NOR or REV positions of the FUNCTION switch as these could cause errors. (see Section IV-2d3).

(g) NOTE:

Trimmer Capacitors C122 and C148 are factory adjusted for optimum performance and should not require readjustment. Adjustment of these trimmers can affect accuracy, therefore, do not attempt to make this adjustment without consulting the factory for advice if it is thought alignment is necessary. Field adjustment is not recommended because special testing equipment and procedures are required.

SECTION VII

Warranty

F. W. BELL, INC. warrants each instrument of its manufacture to be free from defects in material and workmanship. Our obligation under this warranty is limited to servicing or adjusting any instrument returned to our factory for that purpose, and to replacing any defective parts thereof (excluding electron tubes, batteries, transistors and probe sensing elements, which parts are covered by a standard EIA 90-day warranty). This warranty covers instruments which, within one year after delivery to the original purchaser, shall be returned with transportation charges prepaid by the original purchaser, and which upon examination shall disclose to our satisfaction to be defective. If it is determined that the defect has been caused by misuse or abnormal conditions of operation, repairs will be billed at cost after submitting an estimate to the purchaser.

F. W. BELL, INC. reserves the right to make changes in design at any time without incurring any obligation to install same on units previously purchased.

This warranty is expressly in lieu of all other obligations or liabilities on the part of F. W. BELL, INC., and F. W. BELL, INC. neither assumes nor authorizes any other person to assume for them any other liability in connection with the sales of F. W. BELL, INC. instruments.

DAMAGE IN SHIPMENT

The instrument should be examined and tested as soon as it is received. If it does not operate properly, or is damaged in any way, immediately file a claim with the carrier. The claim agent will provide report forms. A copy of the completed form should be forwarded to us. We will then make the necessary arrangements for repair or replacement. All correspondence concerning this instrument should include model and serial numbers.

SHIPPING INSTRUCTIONS

All shipments of F. W. BELL, INC. instruments should be made via Railway Express or Motor Freight. Always use the original shipping carton and inserts, if possible, or pack the instrument in a sturdy container and surround the entire instrument with two or three inches of excelsior or similar shock-absorbing material.

F. W. BELL INC
6120 Hanging Moss Road
Orlando, Florida 32807
Phone: 305-678-6900
TWX: 810-853-3115

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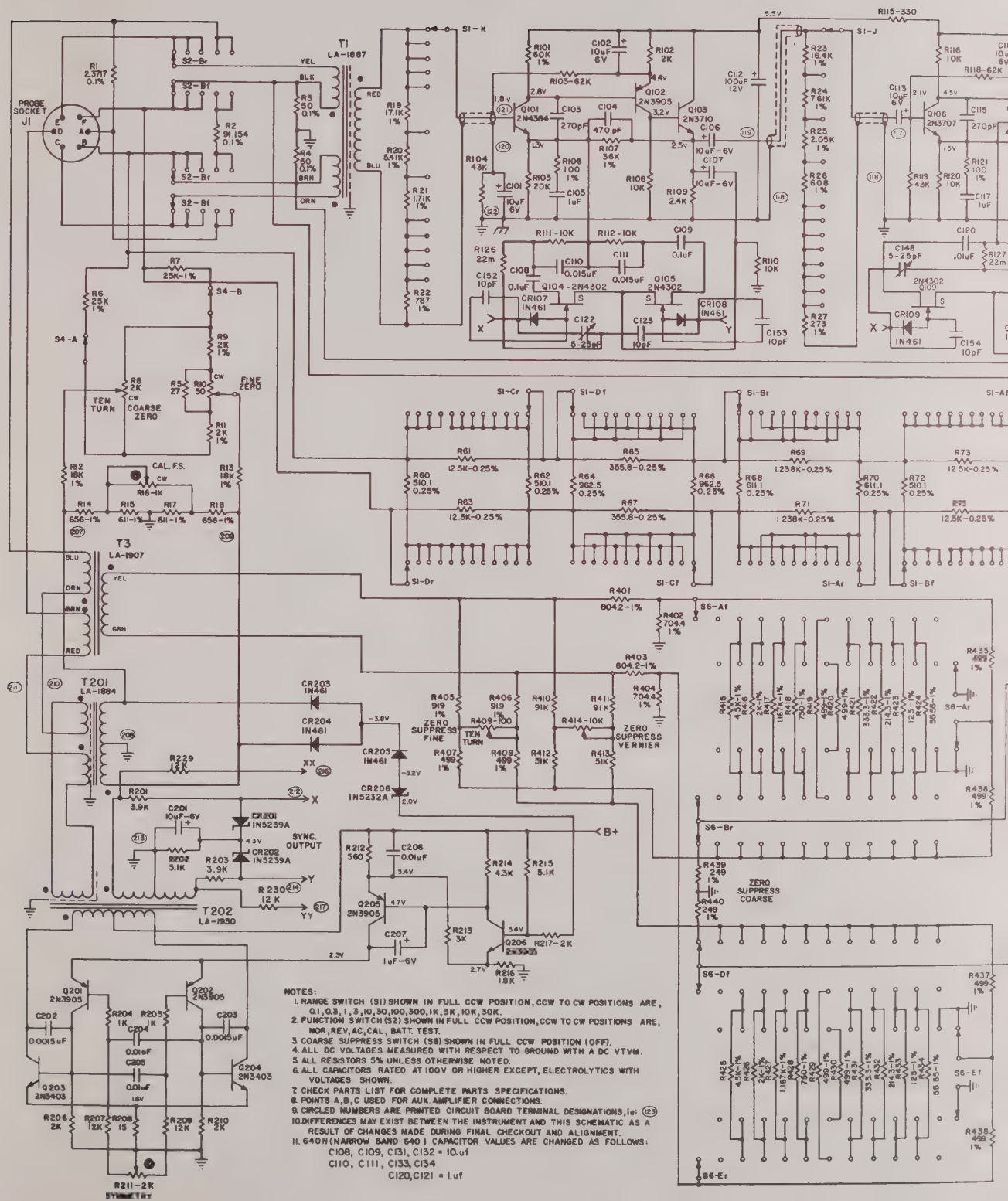
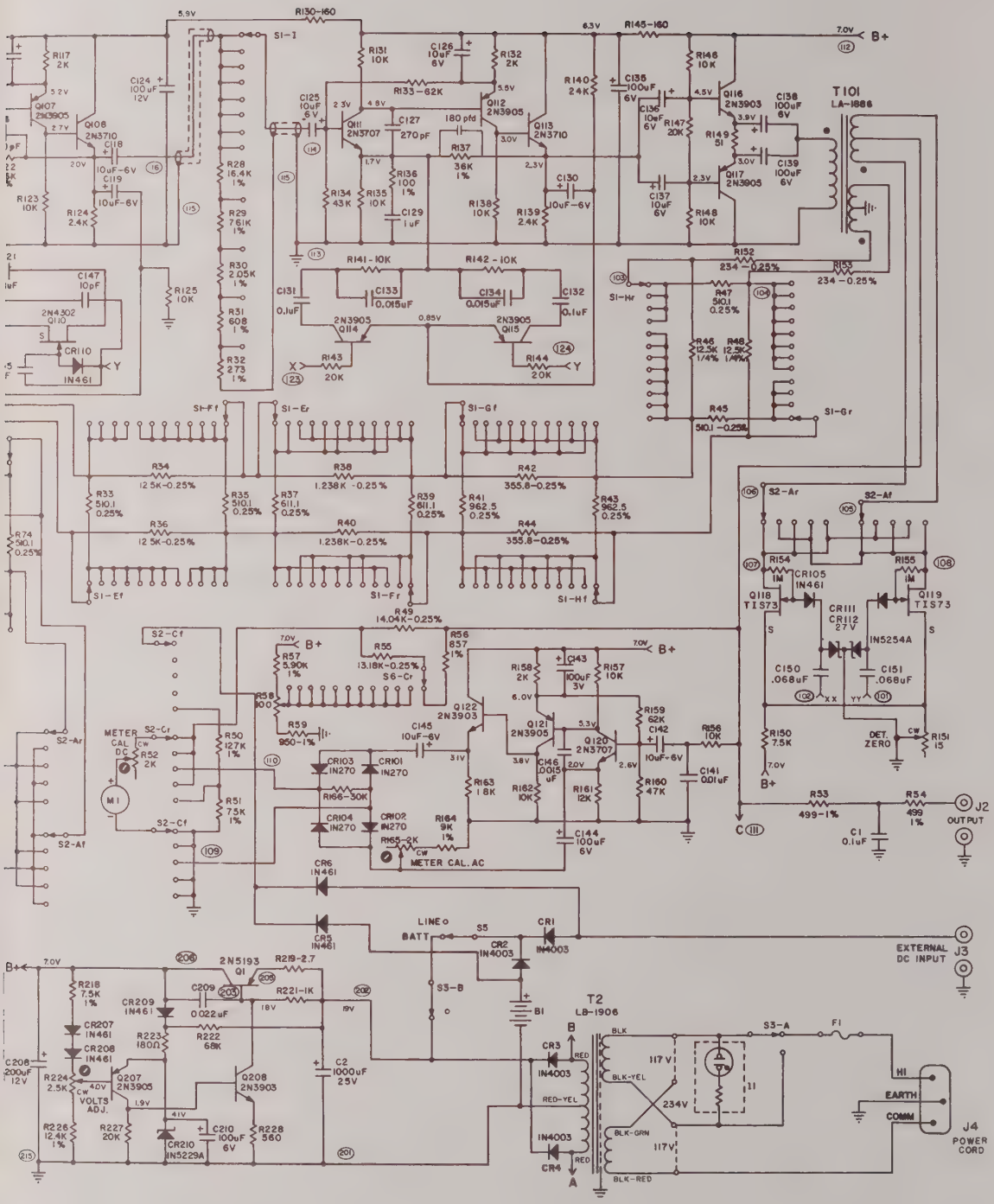
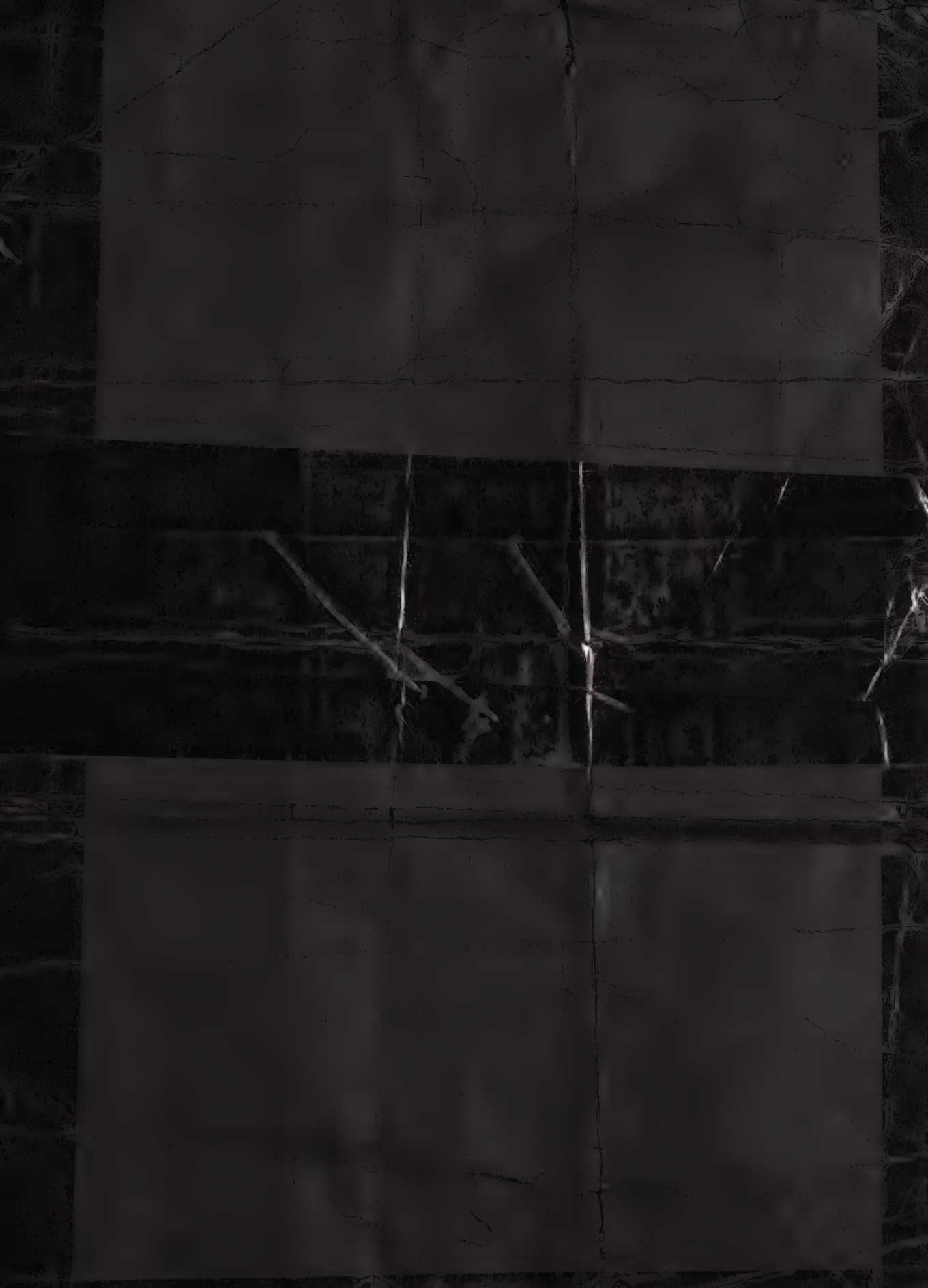
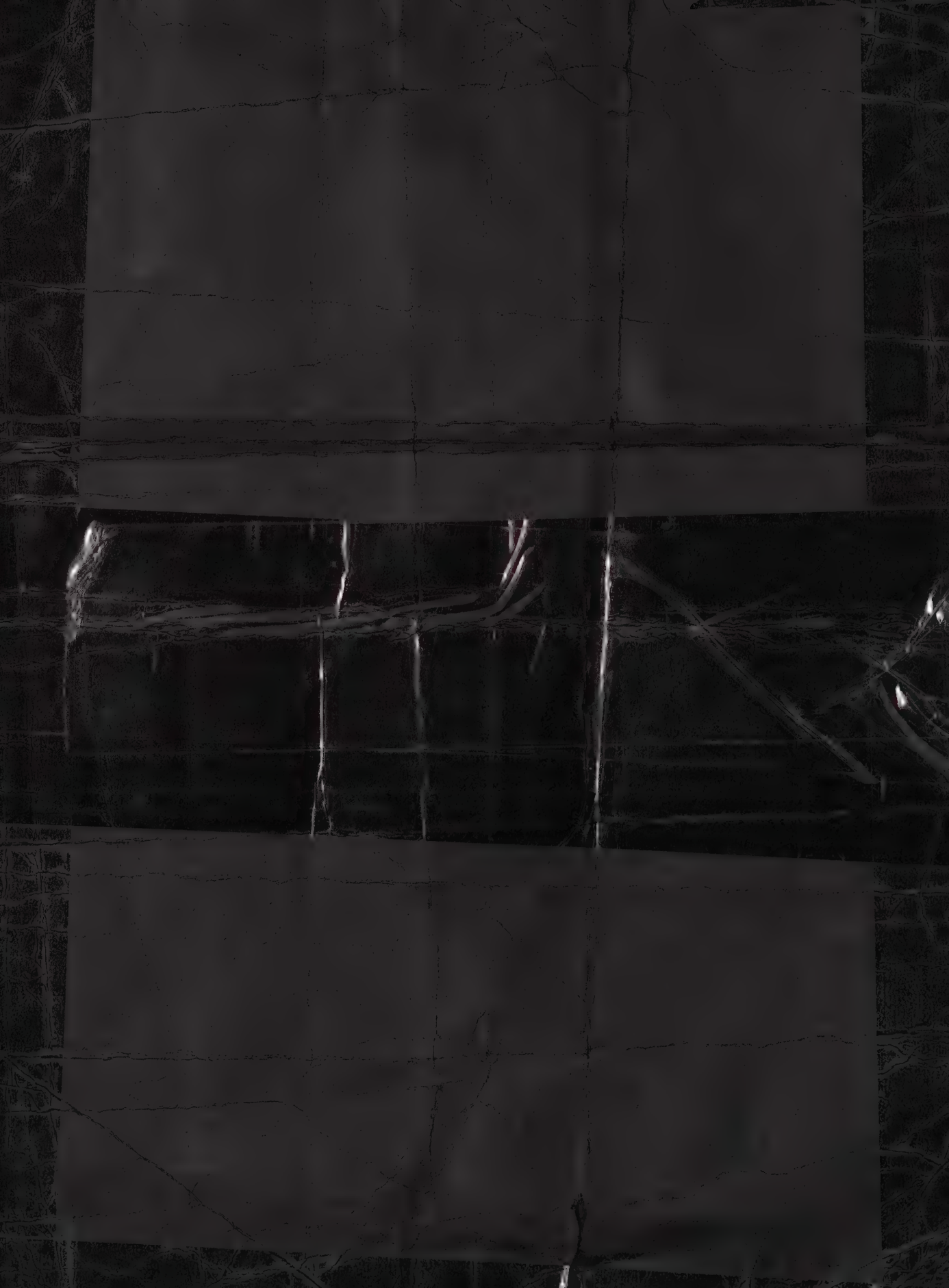

PROBE SOCKET J1
T1 LA-1687
T3 LA-1907
T201 LA-1884
T202 LA-1930
S1-K
S1-J
S1-Cr
S1-Df
S1-Br
S1-Dr
S1-Cf
S1-Ar
S1-Bf
S6-Af
S6-Ar
S6-Br
S6-Df
S6-Ef
S6-Er
TEN TURN COARSE ZERO
FINE ZERO
CAL. F.S.
SYNC. OUTPUT
ZERO SUPPRESS FINE
ZERO SUPPRESS VERNIER
ZERO SUPPRESS COARSE
B+
NOTES:
1. RANGE SWITCH (S1) SHOWN IN FULL CCW POSITION, CCW TO CW POSITIONS ARE, 0.1, 0.3, 1, 3, 10, 30, 100, 300, 1K, 3K, 10K, 30K.
2. FUNCTION SWITCH (S2) SHOWN IN FULL CCW POSITION, CCW TO CW POSITIONS ARE, NOR, REV, AC, CAL, BATT. TEST.
3. COARSE SUPPRESS SWITCH (S6) SHOWN IN FULL CCW POSITION (OFF).
4. ALL DC VOLTAGES MEASURED WITH RESPECT TO GROUND WITH A DC VTVM.
5. ALL RESISTORS 5% UNLESS OTHERWISE NOTED.
6. ALL CAPACITORS RATED AT 100V OR HIGHER EXCEPT, ELECTROLYTICS WITH VOLTAGES SHOWN.
7. CHECK PARTS LIST FOR COMPLETE PARTS SPECIFICATIONS.
8. POINTS A, B, C USED FOR AUX AMPLIFIER CONNECTIONS.
9. CIRCLED NUMBERS ARE PRINTED CIRCUIT BOARD TERMINAL DESIGNATIONS, Ie: (123)
10. DIFFERENCES MAY EXIST BETWEEN THE INSTRUMENT AND THIS SCHEMATIC AS A RESULT OF CHANGES MADE DURING FINAL CHECKOUT AND ALIGNMENT.
11. 640N (NARROW BAND 640) CAPACITOR VALUES ARE CHANGED AS FOLLOWS:
C108, C109, C131, C132 = 10.uf
C110, C111, C133, C134
C120, C121 = 1.uf

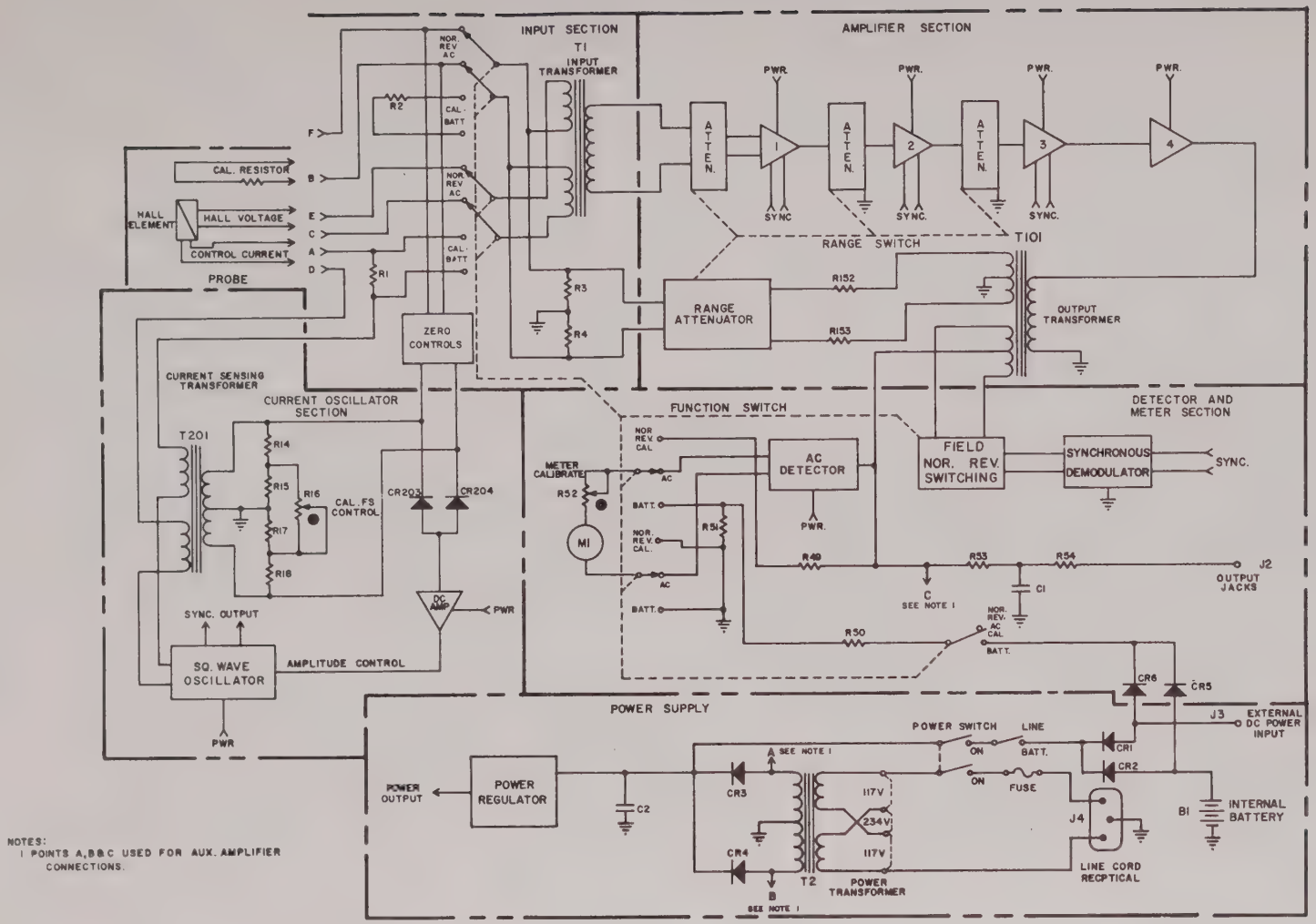


SCHEMATIC DIAGRAM, MODEL 640

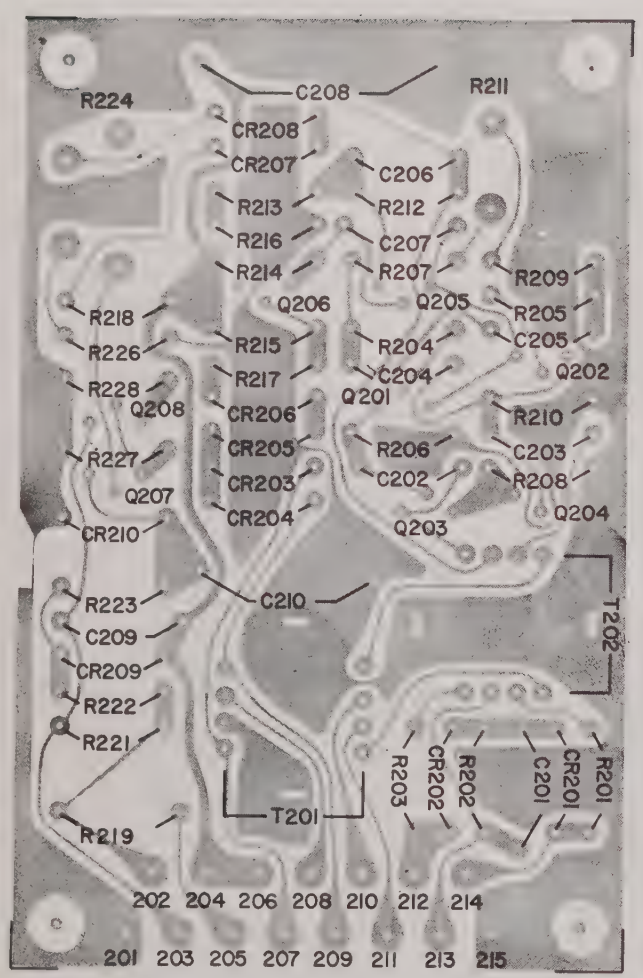
UD-1998



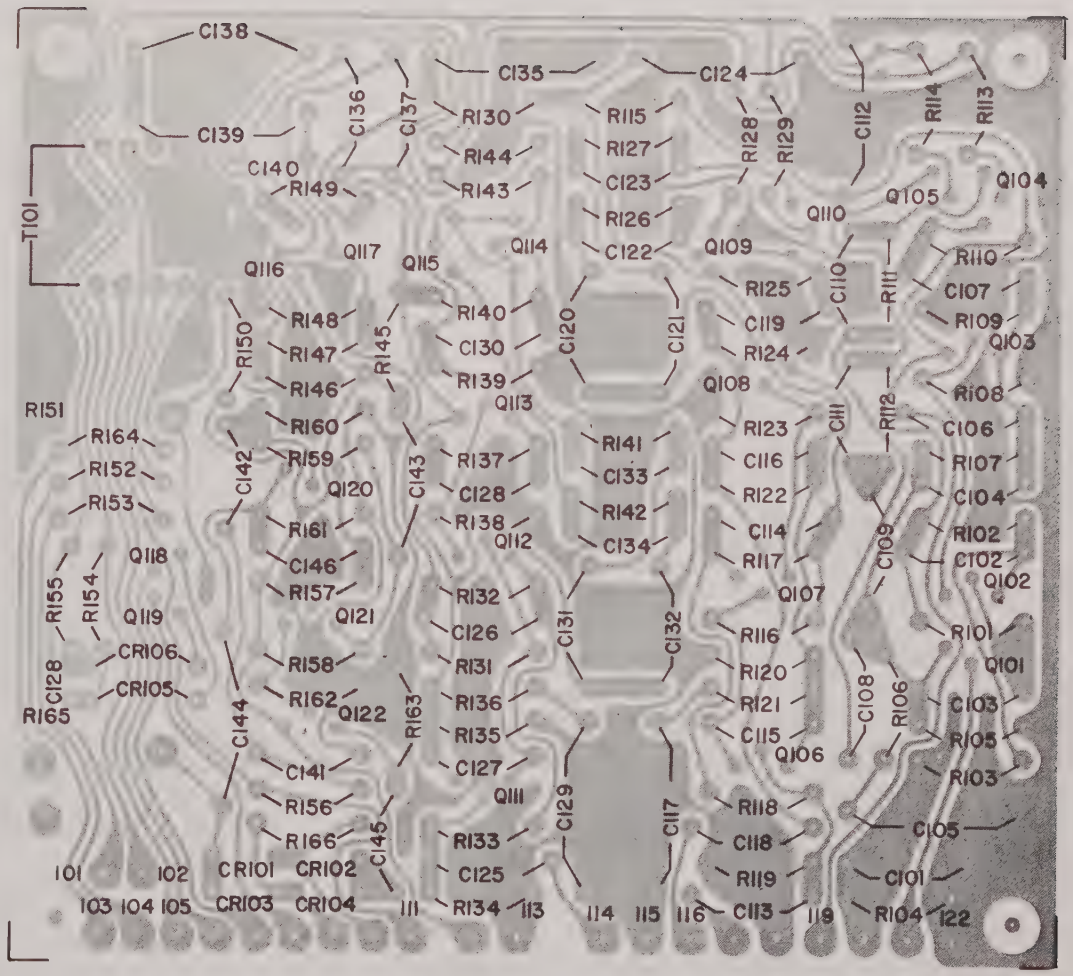




BLOCK DIAGRAM FIG. 4



POWER SUPPLY—CURRENT OSCILLATOR
CIRCUIT BOARD



AMPLIFIER—DETECTOR
CIRCUIT BOARD

ABBREVIATIONS

Cer	Ceramic
Comp	Composition
EC	Electrolytic can type
EM	Electrolytic miniature type
F	Farad
Film	Metal film
Fixed	Fixed
FOP	Factory adjusted for optimum performance
HM	Hot molded
k	Kilo or 10^3
μ	Micro or 10^{-6}
MME	Molded miniature epoxy
Ω	ohm
p	Pico or 10^{-12}
V	Working volts
Var	Variable
W	Watt
WW	Wire wound

PARTS LIST MODEL 640

Schematic Ref.	Value	Spec.		Type		Bell Part No.
R1	2.3717 Ω	1/10 W	1/10%	WW	Fixed	RL-18-2X3717
R2	91.154 Ω	1/10 W	1/10%	WW	Fixed	RL-18-91X154
R3	50 Ω	1/10 W	1/10%	WW	Fixed	RL-18-50
R4	50 Ω	1/10 W	1/10%	WW	Fixed	RL-18-50
R5	27 Ω	½ W	5%	Comp	Fixed	RJ-15-27
R6	25 kΩ	½ W	1%	Film	Fixed	RK-14-25K
R7	25 kΩ	½ W	1%	Film	Fixed	RK-14-25K
R8	2 kΩ			WW	Var	RB-1503-004
R9	2 kΩ	½ W	1%	Film	Fixed	RK-14-2K
R10	50 Ω			HM	Var	RB-505-025
R11	2 kΩ	½ W	1%	Film	Fixed	RK-14-2K
R12	18 kΩ	½ W	1%	Film	Fixed	RK-14-18K
R13	18 kΩ	½ W	1%	Film	Fixed	RK-14-18K
R14	656 Ω	½ W	1%	Film	Fixed	RK-14-656
R15	611 Ω	½ W	1%	Film	Fixed	RK-14-611
R16	1 kΩ			HM	Var	RB-505-024
R17	611 Ω	½ W	1%	Film	Fixed	RK-14-611
R18	656 Ω	½ W	1%	Film	Fixed	RK-14-656
R19	17.1 kΩ	⅛ W	1%	Film	Fixed	RK-33-17X1K
R20	5.41 kΩ	⅛ W	1%	Film	Fixed	RK-33-5X41K
R21	1.71 kΩ	⅛ W	1%	Film	Fixed	RK-33-1X71K
R22	787 Ω	⅛ W	1%	Film	Fixed	RK-33-787
R23	16.4 kΩ	⅛ W	1%	Film	Fixed	RK-33-16X4K
R24	7.61 kΩ	⅛ W	1%	Film	Fixed	RK-33-7X61K
R25	2.08 kΩ	⅛ W	1%	Film	Fixed	RK-33-2X08K
R26	608 Ω	⅛ W	1%	Film	Fixed	RK-33-608
R27	273 Ω	⅛ W	1%	Film	Fixed	RK-33-273
R28	16.4 kΩ	⅛ W	1%	Film	Fixed	RK-33-16X4K
R29	7.61 kΩ	⅛ W	1%	Film	Fixed	RK-33-7X61K
R30	2.05 kΩ	⅛ W	1%	Film	Fixed	RK-33-2X05K
R31	608 Ω	⅛ W	1%	Film	Fixed	RK-33-608
R32	273 Ω	⅛ W	1%	Film	Fixed	RK-33-273
R33	510.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-510X1
R34	12.5 kΩ	⅛ W	¼%	Film	Fixed	RK-35-12X5K
R35	510.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-510X1
R36	12.5 kΩ	⅛ W	¼%	Film	Fixed	RK-35 12X5K
R37	611.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-611X1
R38	1.238 kΩ	⅛ W	¼%	Film	Fixed	RK-35-1X238K
R39	611.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-611X1
R40	1.238 kΩ	⅛ W	¼%	Film	Fixed	RK-35-1X238K
R41	962.5 Ω	⅛ W	¼%	Film	Fixed	RK-35-962X5
R42	355 8 Ω	⅛ W	¼%	Film	Fixed	RK-35-355X8
R43	962.5 Ω	⅛ W	¼%	Film	Fixed	RK-35-962X5
R44	355.8 Ω	⅛ W	¼%	Film	Fixed	RK-35-355X8
R45	510.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-510X1
R46	12.5 kΩ	⅛ W	¼%	Film	Fixed	RK-35-12X5K
R47	510.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-510X1
R48	12.5 kΩ	⅛ W	¼%	Film	Fixed	RK-35-12.5K
R49	14.04 kΩ	½ W	1%	Film	Fixed	RK-14-14X04K
R50	127 kΩ	¼ W	1%	Film	Fixed	RK-23-127K
R51	7.5 kΩ	¼ W	1%	Film	Fixed	RK-23-7X5K
R52	2 kΩ			WW	Var	RS-11-2K
R53	500 Ω	½ W	1%	Film	Fixed	RK-14-500
R54	500 Ω	½ W	1%	Film	Fixed	RK-14-500
R55	13.18 kΩ	⅛ W	¼%	Film	Fixed	RK-35-13X18K
R56	857 Ω	⅛ W	1%	Film	Fixed	RK-33-857
R57	5.90 kΩ	¼ W	1%	Film	Fixed	RK-23-5X90K
R58	100 Ω			WW	Var	RS-11-100
R59	950 Ω	¼ W	1%	Film	Fixed	RK-23-950
R60	510.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-510X1
R61	12.5 kΩ	⅛ W	¼%	Film	Fixed	RK-35-12X5K
R62	510.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-510X1
R63	12.5 kΩ	⅛ W	¼%	Film	Fixed	RK-35-12X5K
R64	962.5 Ω	⅛ W	¼%	Film	Fixed	RK-35-962X5
R65	355.8 Ω	⅛ W	¼%	Film	Fixed	RK-35-355X8
R66	962.5 Ω	⅛ W	¼%	Film	Fixed	RK-35-962X5
R67	355.8 Ω	⅛ W	¼%	Film	Fixed	RK-35-355X8
R68	611.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-611X1
R69	1.238 kΩ	⅛ W	¼%	Film	Fixed	RK-35-1X238K
R70	611.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-611X1
R71	1.238 kΩ	⅛ W	¼%	Film	Fixed	RK-35-1X238K
R72	510.1 Ω	⅛ W	¼%	Film	Fixed	RK-510X1
R73	12X5 kΩ	⅛ W	¼%	Film	Fixed	RK-35-12X5K
R74	510.1 Ω	⅛ W	¼%	Film	Fixed	RK-35-510X1
R401	804.2 Ω	¼ W	1%	Film	Fixed	RK-23-804X2
R402	704.4 Ω	⅛ W	1%	Film	Fixed	RK-33-704X4
R403	804.2 Ω	¼ W	1%	Film	Fixed	RK-23-804X2
R404	704.4 Ω	⅛ W	1%	Film	Fixed	RK-33-704X4
R405	919 Ω	¼ W	1%	Film	Fixed	RK-23-919
R406	919 Ω	¼ W	1%	Film	Fixed	RK-23-919
R407	500 Ω	¼ W	1%	Film	Fixed	RK-23-500
R408	500 Ω	¼ W	1%	Film	Fixed	RK-23-500
R409	120 Ω			WW	Var	RB-1503-005
R410	91 kΩ	½ W	5%	Comp	Fixed	RJ-15-91K
R411	91 kΩ	½ W	5%	Comp	Fixed	RJ-15-91K
R412	51 kΩ	½ W	5%	Comp	Fixed	RJ-15-51K
R413	51 kΩ	½ W	5%	Comp	Fixed	RJ-15-51K
R414	10 kΩ			HM	Var	RB-505-026
R415	4.5 kΩ	⅛ W	1%	Film	Fixed	RK-33-4X5K
R416	2 kΩ	⅛ W	1%	Film	Fixed	RK-33-2K
R417	1.167 kΩ	⅛ W	1%	Film	Fixed	RK-33-1X167K
R418	750 Ω	⅛ W	1%	Film	Fixed	RK-33-750
R419	500 Ω	⅛ W	1%	Film	Fixed	RK-33-500
R420	500 Ω	⅛ W	1%	Film	Fixed	RK-33-500
R421	333.3 Ω	⅛ W	1%	Film	Fixed	RK-33 333X3
R422	214.3 Ω	⅛ W	1%	Film	Fixed	RK-33-214X3
R423	125 Ω	⅛ W	1%	Film	Fixed	RK-33-125
R424	55.55 Ω	⅛ W	1%	Film	Fixed	RK-33-55X55
R425	4.5 kΩ	⅛ W	1%	Film	Fixed	RK-33-4X5K
R426	2 kΩ	⅛ W	1%	Film	Fixed	RK-33-2K

Schematic Ref.	Value	Spec.		Type		Bell Part No.
R427	1.167 kΩ	⅛ W	1%	Film	Fixed	RK-33-1X167K
R428	750 Ω	⅛ W	1%	Film	Fixed	RK-33-750
R429	500 Ω	⅛ W	1%	Film	Fixed	RK-33-500
R430	500 Ω	⅛ W	1%	Film	Fixed	RK-33-500
R431	333.3 Ω	⅛ W	1%	Film	Fixed	RK-33-333X3
R432	214.3 Ω	⅛ W	1%	Film	Fixed	RK-33-214X3
R433	125 Ω	⅛ W	1%	Film	Fixed	RK-33-125
R434	55.55 Ω	⅛ W	1%	Film	Fixed	RK-33-55X55
R435	500 Ω	⅛ W	1%	Film	Fixed	RK-33-500
R436	500 Ω	⅛ W	1%	Film	Fixed	RK-33-500
R437	500 Ω	⅛ W	1%	Film	Fixed	RK-33-500
R438	500 Ω	⅛ W	1%	Film	Fixed	RK-33-500
R439	250 Ω	⅛ W	1%	Film	Fixed	RK-33-250
R440	250 Ω	⅛ W	1%	Film	Fixed	RK-33-250
C1	0.1 μF	200 V		MME	Fixed	CLP-11-X1MF
C2	1000 μF	25 V		EC	Fixed	CHL-01-002

Transistors & Diodes

CR1	1N4003	JT-1N4003
CR2	1N4003	JT-1N4003
CR3	1N4003	JT-1N4003
CR4	1N4003	JT-1N4003
CR5	1N461	JT-1N461
CR6	1N461	JT-1N461
Q1	2N3740	JT-2N3740

Transformers

T1	Input Transformer	LA-1887
T2	Power Transformer	LB-1906
T3	Suppression Transformer	LA-1907

Switches

S1	Range Switch	SC-1845
S2	Function Switch	SA-1848
S3	Power Switch	ST-01-001
S4	Zero Out Switch Part of Fine Zero Control	RB-505-025
S5	Batt-Line Switch	SW-01-002
S6	Coarse Suppression Switch	SB-1849

Meter

M1	0-100 μA Meter	MB-1894

The following parts are on the amplifier circuit board assembly, YP-01-003

R101	10 kΩ	¼ W	1%	Film	Fixed	RK-23-10K
R102	2 kΩ	½ W	5%	Comp	Fixed	RJ-15-2K
R103	62 kΩ	½ W	5%	Comp	Fixed	RJ-15-62K
R104	43 kΩ	½ W	5%	Comp	Fixed	RJ-15-43K
R105	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R106	100 Ω	¼ W	1%	Film	Fixed	RK-23-100
R107	36 kΩ	¼ W	1%	Film	Fixed	RK-23-36K
R108	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R109	2.4 kΩ	½ W	5%	Comp	Fixed	RJ-15-2X4K
R110	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R111	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R112	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R115	330 Ω	½ W	5%	Comp	Fixed	RJ-15-330
R116	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R117	2 kΩ	½ W	5%	Comp	Fixed	RJ-15-2K
R118	62 kΩ	½ W	5%	Comp	Fixed	RJ-15-62K

Schematic Ref.	Value	Spec.		Type		Bell Part No.
R119	43 kΩ	½ W	5%	Comp	Fixed	RJ-15-43K
R120	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R121	100 Ω	¼ W	1%	Film	Fixed	RK-23-100
R122	36 kΩ	¼ W	1%	Film	Fixed	RK-23-36K
R123	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R124	2.4 kΩ	½ W	5%	Comp	Fixed	RJ-15-2X4K
R125	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R126	22 MΩ	½ W	5%	Comp	Fixed	RJ-15-22M
R127	22 MΩ	½ W	5%	Comp	Fixed	RJ-15-22M
R130	160 Ω	½ W	5%	Comp	Fixed	RJ-15-160
R131	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R132	2 kΩ	½ W	5%	Comp	Fixed	RJ-15-2K
R133	62 kΩ	½ W	5%	Comp	Fixed	RJ-15-62K
R134	43 kΩ	½ W	5%	Comp	Fixed	RJ-15-43K
R135	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R136	100 Ω	¼ W	1%	Film	Fixed	RK-23-100
R137	36 kΩ	¼ W	1%	Film	Fixed	RK-23-36K
R138	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R139	2.4 kΩ	½ W	5%	Comp	Fixed	RJ-15-2X4K
R140	24 kΩ	½ W	5%	Comp	Fixed	RJ-15-24K
R141	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R142	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R143	20 kΩ	½ W	5%	Comp	Fixed	RJ-15-20K
R144	20 kΩ	½ W	5%	Comp	Fixed	RJ-15-20K
R145	160 Ω	½ W	5%	Comp	Fixed	RJ-15-160
R146	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R147	20 kΩ	½ W	5%	Comp	Fixed	RJ-15-20K
R148	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R149	51 Ω	½ W	5%	Comp	Fixed	RJ-15-51
R150	7.5 kΩ	½ W	5%	Comp	Fixed	RJ-15-7X5K
R151	15 Ω			WW	Var	RS-12-15
R152	234 Ω	¼ W	¼%	Film	Fixed	RK-25-234
R153	234 Ω	¼ W	¼%	Film	Fixed	RK-25-234
R154	IM	½ W	5%	Comp	Fixed	RJ-15-IM
R155	IM	½ W	5%	Comp	Fixed	RJ-15-IM
R156	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R157	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R158	2 kΩ	½ W	5%	Comp	Fixed	RJ-15-2K
R159	62 kΩ	½ W	5%	Comp	Fixed	RJ-15-62K
R160	47 kΩ	½ W	5%	Comp	Fixed	RJ-15-47K
R161	12 kΩ	½ W	5%	Comp	Fixed	RJ-15-12K
R162	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R163	1.8 kΩ	½ W	5%	Comp	Fixed	RJ-15-1X8K
R164	9 kΩ	¼ W	1%	Film	Fixed	RK-23-9K
R165	2 kΩ			WW	Var	RS-12-2K
R166	30 kΩ	½ W	5%	Comp	Fixed	RJ-15-30K
C101	10 μF	6 V		EM	Fixed	CFH-11-10MF
C102	10 μF	6 V		EM	Fixed	CFH-11-10MF
C103	270 pF	1 kV		Cer	Fixed	CRK-03-270PF
C104	470 pF	1 kV		Cer	Fixed	CRK-03-470PF
C105	1 μF	100 V		MME	Fixed	CMK-02-1MF
C106	10 μF	6 V		EM	Fixed	CFH-11-10MF
C107	10 μF	6 V		EM	Fixed	CFH-11-10MF
C108	0.1 μF	200 V		MME	Fixed	CLP-11-X1MF
C109	0.1 μF	200 V		MME	Fixed	CLP-11-X1MF
C110	0.015 μF	200 V		MME	Fixed	CLP-11-X015MF
C111	0.015 μF	200 V		MME	Fixed	CLP-11-X015MF
C112	100 μF	12 V		EM	Fixed	CFH-32-100MF
C113	10 μF	6 V		EM	Fixed	CFH-11-10MF
C114	10 μF	6 V		EM	Fixed	CFH-11-10MF
C115	270 pF	1 kV		Cer	Fixed	CRK-03-270PF
C116	470 pF	1 kV		Cer	Fixed	CRK-03-470PF
C117	1 μF	100 V		MME	Fixed	CMK-02-1MF
C118	10 μF	6 V		EM	Fixed	CFH-11-10MF
C119	10 μF	6 V		EM	Fixed	CFH-11-10MF

Schematic Ref.	Value	Spec.	Type		Bell Part No.
C120	0.01 μF	200 V	MME	Fixed	CLP-11-X01MF
C121	0.01 μF	200 V	MME	Fixed	CLP-11-X01MF
C122	5-25 pF	600 V	Cer	Var	CV-03-5-25PF
C123	10 pF	1000 V	Cer	Fixed	CRK-03-10PF
C124	100 μF	12 V	EM	Fixed	CFH-32-100MF
C125	10 μF	6 V	EM	Fixed	CFH-11-10MF
C126	10 μF	6 V	EM	Fixed	CFH-11-10MF
C127	270 pF	1 kV	Cer	Fixed	CRK-03-270PF
C129	1 μF	100 V	MME	Fixed	CMK-02-1MF
C130	10 μF	6 V	EM	Fixed	CFH-11-10MF
C131	0.1 μF	200 V	MME	Fixed	CLP-11-X1MF
C132	0.1 μF	200 V	MME	Fixed	CLP-11-X1MF
C133	0.015 μF	200 V	MME	Fixed	CLP-11-X015MF
C134	0.015 μF	200 V	MME	Fixed	CLP-11-X015MF
C135	100 μF	12 V	EM	Fixed	CFH-32-100MF
C136	10 μF	6 V	EM	Fixed	CFH-11-10MF
C137	10 μF	6 V	EM	Fixed	CFH-11-10MF
C138	100 μF	6 V	EM	Fixed	CFH-11-100MF
C139	100 μF	6 V	EM	Fixed	CFH-11-100MF
C140	220 pF	1 kV	Cer	Fixed	CRK-03-220PF
C141	0.01 μF	200 V	MME	Fixed	CLP-11-X01MF
C142	10 μF	6 V	EM	Fixed	CFH-11-10MF
C143	100 μF	3 V	EM	Fixed	CFH-01-100MF
C144	100 μF	6 V	EM	Fixed	CFH-11-100MF
C145	10 μF	6 V	EM	Fixed	CFH-11-10MF
C146	0.0015 μF	200 V	MME	Fixed	CLP-11-X0015MF
C147	10 pF	200 V	Cer	Fixed	CRK-03-10PF
C148	5-25 pF	600 V	Cer	Var	CV-03-5-25PF
C150	.068 F	200 V	MME		CLP-11-X068MF
C151	.068 F	200 V	MME		CLP-11-X068MF
C152	10 pF	1 kV	Cer	Fixed	CRK-03-10PF
C153	10 pF	1 kV	Cer	Fixed	CRK-03-10PF
C154	10 pF	1 kV	Cer	Fixed	CRK-03-10PF
C155	10 pF	1 kV	Cer	Fixed	CRK-03-10PF

Transistors & Diodes

CR101	1N270	JT-1N270
CR102	1N270	JT-1N270
CR103	1N270	JT-1N270
CR104	1N270	JT-1N270
CR105	1N461	JT-1N461
CR106	1N461	JT-1N461
CR107	1N461	JT-1N461
CR108	1N461	JT-1N461
CR109	1N461	JT-1N461
CR110	1N461	JT-1N461
CR111	1N5254A	JT-1N5254A
CR112	1N5254A	JT-1N5254A
Q101	2N4384	JT-2N4384
Q102	2N3905	JT-2N3905
Q103	2N3710	JT-2N3710
Q104	2N4302	JT-2N4302
Q105	2N4302	JT-2N4302
Q106	2N3707	JT-2N3707
Q107	2N3905	JT-2N3905
Q108	2N3710	JT-2N3710
Q109	2N4302	JT-2N4302
Q110	2N4302	JT-2N4302
Q111	2N3707	JT-2N3707
Q112	2N3905	JT-2N3905
Q113	2N3710	JT-2N3710
Q114	2N3905	JT-2N3905
Q115	2N3905	JT-2N3905
Q116	2N3903	JT-2N3903
Q117	2N3905	JT-2N3905
Q118	T1S73	JK-01-005
Q119	T1S73	JK-01-005
Q120	2N3707	JT-2N3707
Q121	2N3905	JT-2N3905
Q122	2N3903	JT-2N3903

The following parts are on the power supply — current oscillator circuit board assembly YP-01-004

Schematic Ref.	Value	Spec.	Type		Bell Part No.
R201	3.9 kΩ	½ W 5%	Comp	Fixed	RJ-15-3X9K
R202	5.1 kΩ	½ W 5%	Comp	Fixed	RJ-15-5X1K
R203	3.9 kΩ	½ W 5%	Comp	Fixed	RJ-15-3X9K
R204	1 kΩ	½ W 5%	Comp	Fixed	RJ-15-1K
R205	1 kΩ	½ W 5%	Comp	Fixed	RJ-15-1K
R206	2 kΩ	½ W 5%	Comp	Fixed	RJ-15-2K
R207	12 kΩ	½ W 5%	Comp	Fixed	RJ-15-12K
R208	15 Ω	½ W 5%	Comp	Fixed	RJ-15-15
R209	12 kΩ	½ W 5%	Comp	Fixed	RJ-15-12K
R210	2 kΩ	½ W 5%	Comp	Fixed	RJ-15-2K
R211	2 kΩ		WW	Var	RS-11-2K
R212	560 Ω	½ W 5%	Comp	Fixed	RJ-15-560
R213	3 kΩ	½ W 5%	Comp	Fixed	RJ-15-3K
R214	4.3 kΩ	½ W 5%	Comp	Fixed	RJ-15-4X3K
R215	5.1 kΩ	½ W 5%	Comp	Fixed	RJ-15-5X1K
R216	1.8 kΩ	½ W 5%	Comp	Fixed	RJ-15-1X8K
R217	2 kΩ	½ W 5%	Comp	Fixed	RJ-15-2K
R218	7.5 kΩ	¼ W 1%	Film	Fixed	RK-23-7X5K
R219	2.7 Ω	2 W 5%	WW	Fixed	RL-13-2X7
R221	1 kΩ	½ W 5%	Comp	Fixed	RJ-15-1K
R222	68 kΩ	½ W 5%	Comp	Fixed	RJ-15-68K
R223	510 Ω	½ W 5%	Comp	Fixed	RJ-15-510
R224	2.5 kΩ		WW	Var	RS-11-2X5K
R225	110 Ω	½ W 5%	Comp	Fixed	RJ-15-110
R226	16 kΩ	¼ W 1%	Film	Fixed	RK-23-16K
R227	20 kΩ	½ W 5%	Comp	Fixed	RJ-15-20K
R228	560 Ω	½ W 5%	Comp	Fixed	RJ-15-560
C201	10 μF	6 V	EM	Fixed	CFH-11-10MF
C202	0.0015 μF	200 V	MME	Fixed	CLP-11-X0015MF
C203	0.0015 μF	200 V	MME	Fixed	CLP-11-X0015MF
C204	0.01 μF	200 V	MME	Fixed	CLP-11-X01MF
C205	0.01 μF	200 V	MME	Fixed	CLP-11-X01MF
C206	0.01 μF	200 V	MME	Fixed	CLP-11-X01MF
C207	1 μF	6 V	EM	Fixed	CFH-11-1MF
C208	200 μF	12 V	EM	Fixed	CFH-32-200MF
C209	0.022 μF	200 V	MME	Fixed	CLP-11-X022MF
C210	100 μF	6 V	EM	Fixed	CFH-11-100MF

Transistors & Diodes

CR201	1N5239A	JT-1N5239A
CR202	1N5239A	JT-1N5239A
CR203	1N461	JT-1N461
CR204	1N461	JT-1N461
CR205	1N461	JT-1N461
CR206	1N5232A	JT-1N5232A
CR207	1N461	JT-1N461
CR208	1N461	JT-1N461
CR209	1N461	JT-1N461
CR210	1N5229A	JT-1N5229A
Q201	2N3905	JT-2N3905
Q202	2N3905	JT-2N3905
Q203	2N3403	JT-2N3403
Q204	2N3403	JT-2N3403
Q205	2N3905	JT-2N3905
Q206	2N3903	JT-2N3903
Q207	2N3905	JT-2N3905
Q208	2N3903	JT-2N3903

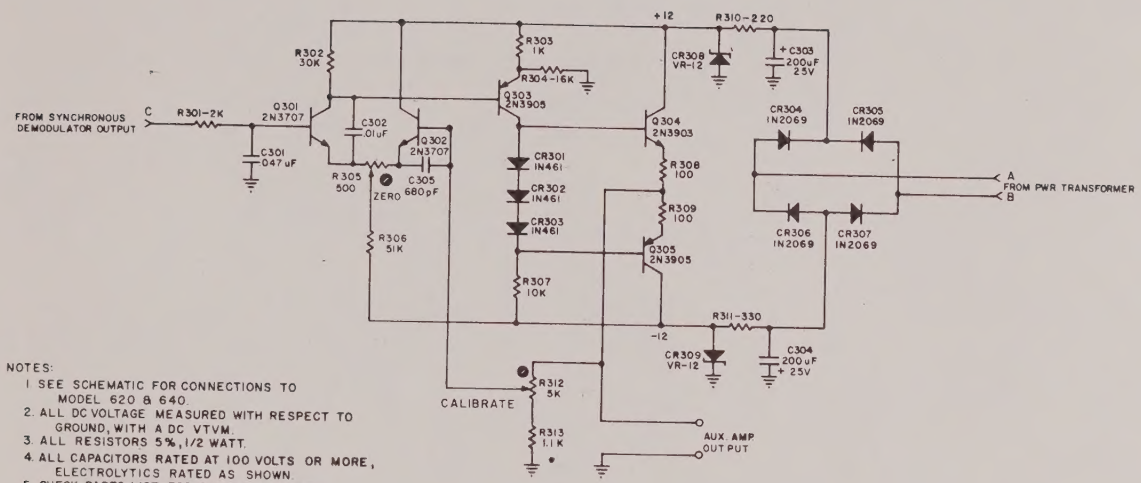
Transformer

T101	Output Transformer	LA-1886

Transformers

T201	Current Sensing Transformer	LA-1884
T202	Current Output Transformer	LA-1930

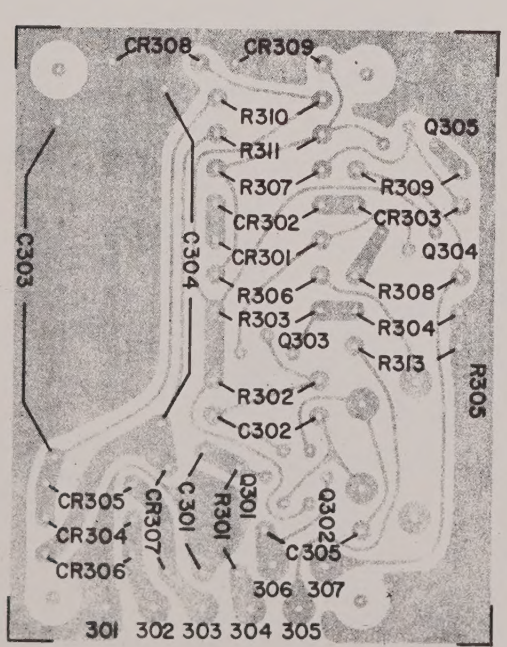
AUXILIARY AMPLIFIER



NOTES:

1. SEE SCHEMATIC FOR CONNECTIONS TO MODEL 620 & 640.
2. ALL DC VOLTAGE MEASURED WITH RESPECT TO GROUND, WITH A DC VTVM.
3. ALL RESISTORS 5%, 1/2 WATT.
4. ALL CAPACITORS RATED AT 100 VOLTS OR MORE, ELECTROLYTICS RATED AS SHOWN.
5. CHECK PARTS LIST FOR COMPLETE PARTS SPECIFICATIONS.

SCHEMATIC DIAGRAM



CIRCUIT BOARD

ABBREVIATIONS

Cer	Ceramic
Comp	Composition
EC	Electrolytic can type
EM	Electrolytic miniature type
F	Farad
Film	Metal film
Fixed	Fixed
FOP	Factory adjusted for optimum performance
HM	Hot molded

k	Kilo or 10^3
μ	Micro or 10^{-6}
MME	Molded miniature epoxy
Ω	ohm
p	Pico or 10^{-12}
V	Working volts
Var	Variable
W	Watt
WW	Wire wound

PARTS LIST AUXILIARY AMPLIFIER

Schematic Ref.	Value	Spec.		Type		Bell Part No.
R301	2 kΩ	½ W	5%	Comp	Fixed	RJ-15-2K
R302	30 kΩ	½ W	5%	Comp	Fixed	RJ-15-30K
R303	1 kΩ	½ W	5%	Comp	Fixed	RJ-15-1K
R304	16 kΩ	½ W	5%	Comp	Fixed	RJ-15-16K
R305	500 Ω			WW	Var	RS-11-500
R306	51 kΩ	½ W	5%	Comp	Fixed	RJ-15-51K
R307	10 kΩ	½ W	5%	Comp	Fixed	RJ-15-10K
R308	100 Ω	½ W	5%	Comp	Fixed	RJ-15-100
R309	100 Ω	½ W	5%	Comp	Fixed	RJ-15-100
R310	220 Ω	½ W	5%	Comp	Fixed	RJ-15-220
R311	330 Ω	½ W	5%	Comp	Fixed	RJ-15-330
R312	5 kΩ			HM	Var	RB-505-003
R313	1.1 kΩ	½ W	5%	Comp	Fixed	RJ-15-1X1K
C301	0.047 μF	200 V		MME	Fixed	CLP-11-X047MF
C302	0.01 μF	200 V		MME	Fixed	CLP-11-X01MF
C303	200 μF	25 V		EM	Fixed	CFH-57-200MF
C304	200 μF	25 V		EM	Fixed	CFH-57-200MF
CR301		1N461				JT-1N461
CR302		1W461				JT-1N461
CR303		1N461				JT-1W461
CR304		1N2069 or 1N4003				JT-1N2069
CR305		1N2069 or 1N4003				JT-1N2069
CR306		1N2069 or 1N4003				JT-1N2069
CR307		1N2069 or 1N4003				JT-1N2069
CR308		VR-12				JF-01-001
CR309		VR-12				JF-01-001
Q301		2N3707				JT-2N3707
Q302		2N3707				JT-2N3707
Q303		2N3905				JT-2N3905
Q304		2N3903				JT-2N3903
Q305		2N3905				JT-2N3905

MP MASTER
MA30392

SMEAD–HASTINGS, MN
NO. T118W